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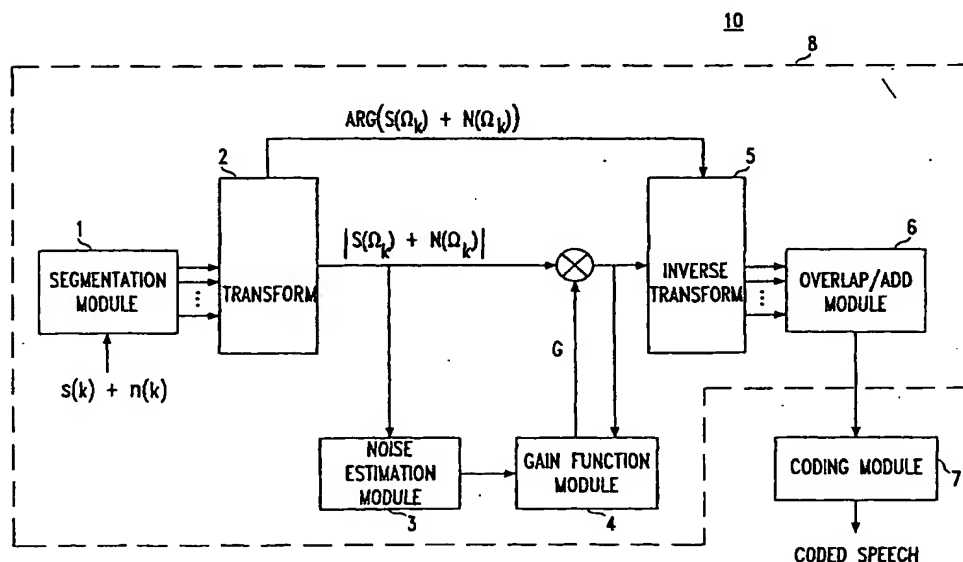
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(54) Title: SPEECH ENHANCEMENT WITH GAIN LIMITATIONS BASED ON SPEECH ACTIVITY



(57) Abstract: An apparatus and method for data processing that improves estimation of spectral parameters of speech data and reduces algorithmic delay in a data coding operation. Estimation of spectral parameters is improved by adaptively adjusting a gain function used to enhance data based on whether the data contains information speech and noise or noise only. Delay is reduced by extracting coding parameters using incompletely processed data.

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SPEECH ENHANCEMENT WITH GAIN LIMITATIONS BASED ON SPEECH ACTIVITY

Cross-Reference to Related Applications

5 This application claims the benefit of the filing date of U.S. Provisional Application No. 60/119,279, filed February 9, 1999, and is incorporated herein by reference.

Field of the Invention

10 This invention relates to enhancement processing for speech coding (*i.e.*, speech compression) systems, including low bit-rate speech coding systems such as MELP.

Background of the Invention

15 Low bit-rate speech coders, such as parametric speech coders, have improved significantly in recent years. However, low-bit rate coders still suffer from a lack of robustness in harsh acoustic environments. For example, artifacts introduced by low bit-rate parametric coders in medium and low signal-to-noise ratio (SNR) conditions can affect intelligibility of coded speech.

20 Tests show that significant improvements in coded speech can be made when a low bit-rate speech coder is combined with a speech enhancement preprocessor. Such enhancement preprocessors typically have three main components: a spectral analysis/synthesis system (usually realized by a windowed fast Fourier transform/inverse fast Fourier transform (FFT/IFFT), a noise estimation process, and a spectral gain computation. The noise estimation process typically involves some type of voice activity detection or spectral
25 minimum tracking technique. The computed spectral gain is applied only to the Fourier magnitudes of each data frame (*i.e.*, segment) of a speech signal. An example of a speech enhancement preprocessor is provided in Y. Ephraim et al., "Speech Enhancement Using a Minimum Mean-Square Error Log-Spectral

Amplitude Estimator," IEEE Trans. Acoustics, Speech and Signal Processing, Vol. 33, pp. 443-445, April 1985, which is hereby incorporated by reference in its entirety. As is conventional, the spectral gain comprises individual gain values to be applied to the individual subbands output by the FFT process.

5 A speech signal may be viewed as representing periods of articulated speech (that is, periods of "speech activity") and speech pauses. A pause in articulated speech results in the speech signal representing background noise only, while a period of speech activity results in the speech signal representing both articulated speech and background noise. Enhancement
10 preprocessors function to apply a relatively low gain during periods of speech pauses (since it is desirable to attenuate noise) and a higher gain during periods of speech (to lessen the attenuation of what has been articulated). However, switching from a low to a high gain value to reflect, for example, the onset of speech activity after a pause, and *vice-versa*, can result in structured "musical"
15 (or "tonal") noise artifacts which are displeasing to the listener. In addition, enhancement preprocessors themselves can introduce degradations in speech intelligibility as can speech coders used with such preprocessors.

 To address the problem of structured musical noise, some enhancement preprocessors uniformly limit the gain values applied to all data frames of the
20 speech signal. Typically, this is done by limiting an "*a priori*" signal to noise ratio (SNR) which is a functional input to the computation of the gain. This limitation on gain prevents the gain applied in certain data frames (such as data frames corresponding to speech pauses) from dropping too low and contributing to significant changes in gain between data frames (and thus, structured musical
25 noise). However, this limitation on gain does not adequately ameliorate the intelligibility problem introduced by the enhancement preprocessor or the speech coder.

Summary of the Invention

The present invention overcomes the problems of the prior art to both limit structured musical noise and increase speech intelligibility. In the context of an enhancement preprocessor, an illustrative embodiment of the invention makes a determination of whether the speech signal to be processed represents articulated speech or a speech pause and forms a unique gain to be applied to the speech signal. The gain is unique in this context because the lowest value the gain may assume (*i.e.*, its lower limit) is determined based on whether the speech signal is known to represent articulated speech or not. In accordance with this embodiment, the lower limit of the gain during periods of speech pause is constrained to be higher than the lower limit of the gain during periods of speech activity.

In the context of this embodiment, the gain that is applied to a data frame of the speech signal is adaptively limited based on limited *a priori* SNR values. These *a priori* SNR values are limited based on (a) whether articulated speech is detected in the frame and (b) a long term SNR for frames representing speech. A voice activity detector can be used to distinguish between frames containing articulated speech and frames that contain speech pauses. Thus, the lower limit of *a priori* SNR values may be computed to be a first value for a frame representing articulated speech and a different second value, greater than the first value, for a frame representing a speech pause. Smoothing of the lower limit of the *a priori* SNR values is performed using a first order recursive system to provide smooth transitions between active speech and speech pause segments of the signal.

An embodiment of the invention may also provide for reduced delay of coded speech data that can be caused by the enhancement preprocessor in combination with a speech coder. Delay of the enhancement preprocessor and coder can be reduced by having the coder operate, at least partially, on incomplete data samples to extract at least some coder parameters. The total delay imposed by the preprocessor and coder is usually equal to the sum of the

delay of the coder and the length of overlapping portions of frames in the enhancement preprocessor. However, the invention takes advantage of the fact that some coders store "look-ahead" data samples in an input buffer and use these samples to extract coder parameters. The look-ahead samples typically have less influence on the quality of coded speech than other samples in the input buffer. Thus, in some cases, the coder does not need to wait for a fully processed, *i.e.*, complete, data frame from the preprocessor, but instead can extract coder parameters from incomplete data samples in the input buffer. By operating on incomplete data samples, delay of the enhancement preprocessor and coder can be reduced without significantly affecting the quality of the coded data.

For example, delay in a speech preprocessor and speech coder combination can be reduced by multiplying an input frame by an analysis window and enhancing the frame in the enhancement preprocessor. After the frame is enhanced, the left half of the frame is multiplied by a synthesis window and the right half is multiplied by an inverse analysis window. The synthesis window can be different from the analysis window, but preferably is the same as the analysis window. The frame is then added to the speech coder input buffer, and coder parameters are extracted using the frame. After coder parameters are extracted, the right half of the frame in the speech coder input buffer is multiplied by the analysis and the synthesis window, and the frame is shifted in the input buffer before the next frame is input. The analysis windows, and synthesis window used to process the frame in the coder input buffer can be the same as the analysis and synthesis windows used in the enhancement preprocessor, or can be slightly different, *e.g.*, the square root of the analysis window used in the preprocessor. Thus, the delay imposed by the preprocessor can be reduced to a very small level, *e.g.*, 1-2 milliseconds.

These and other aspects of the invention will be appreciated and/or obvious in view of the following description of the invention.

Brief Description of the Drawings

The invention is described in connection with the following drawings where reference numerals indicate like elements and wherein:

5 Figure 1 is a schematic block diagram of an illustrative embodiment of the invention.

 Figure 2 is a flowchart of steps for a method of processing speech and other signals in accordance with the embodiment of Figure 1.

 Figure 3 is a flowchart of steps for a method for enhancing speech signals in accordance with the embodiment of Figure 1.

10 Figure 4 is a flowchart of steps for a method of adaptively adjusting an *a priori* SNR value in accordance with the embodiment of Figure 1.

 Figure 5 is a flowchart of the steps for a method of applying a limit to the *a priori* signal to noise ratio for use in a gain computation.

15 **Detailed Description**

A. Introduction to Illustrative Embodiments

20 As is conventional in the speech coding art, the illustrative embodiment of the present invention is presented as comprising individual functional blocks (or "modules"). The functions these blocks represent may be provided through the use of either shared or dedicated hardware, including, but not limited to, hardware capable of executing software. For example, the functions of blocks 1-5 presented in Figure 1 may be provided by a single shared processor. (Use of the term "processor" should not be construed to refer exclusively to hardware capable of executing software.)

Illustrative embodiments may be realized with digital signal processor (DSP) or general purpose personal computer (PC) hardware, available from any of a number of manufacturers, read-only memory (ROM) for storing software performing the operations discussed below, and random access memory (RAM) for storing DSP/PC results. Very large scale integration (VLSI) hardware embodiments, as well as custom VLSI circuitry in combination with a general purpose DSP/PC circuit, may also be provided.

Illustrative software for performing the functions presented in Figure 1 is provided in the Software Appendix hereto.

B. The Illustrative Embodiment

Figure 1 presents a schematic block diagram of an illustrative embodiment 8 of the invention. As shown in Figure 1, the illustrative embodiment processes various signals representing speech information. These signals include a speech signal (which includes a pure speech component, $s(k)$, and a background noise component, $n(k)$), data frames thereof, spectral magnitudes, spectral phases, and coded speech. In this example, the speech signal is enhanced by a speech enhancement preprocessor 8 and then coded by a coder 7. The coder 7 in this illustrative embodiment is a 2400 bps MIL Standard MELP coder, such as that described in A. McCree et al., "A 2.4 KBIT/S MELP Coder Candidate for the New U.S. Federal Standard," Proc., IEEE Intl. Conf. Acoustics, Speech, Signal Processing (ICASSP), pp. 200-203, 1996, which is hereby incorporated by reference in its entirety. Figures 2, 3, 4, and 5 present flow diagrams of the processes carried out by the modules presented in Figure 1.

1. The Segmentation Module

The speech signal, $s(k) + n(k)$, is input into a segmentation module 1. The segmentation module 1 segments the speech signal into frames of 256 samples of speech and noise data (see step 100 of Figure 2; the size of the data frame can be any desired size, such as the illustrative 256 samples), and applies an analysis window to the frames prior to transforming the frames into the

frequency domain (see step 200 of Figure 2). As is well known, applying the analysis window to the frame affects the spectral representation of the speech signal.

The analysis window is tapered at both ends to reduce cross talk between subbands in the frame. Providing a long taper for the analysis window significantly reduces cross talk, but can result in increased delay of the preprocessor and coder combination 10. The delay inherent in the preprocessing and coding operations can be minimized when the frame advance (or a multiple thereof) of the enhancement preprocessor 8 matches the frame advance of the coder 7. However, as the shift between later synthesized frames in the enhancement preprocessor 8 increases from the typical half-overlap (e.g., 128 samples) to the typical frame shift of the coder 7 (e.g., 180 samples), transitions between adjacent frames of the enhanced speech signal $\hat{s}(k)$ become less smooth. These discontinuities arise because the analysis window attenuates the input signal most at the edges of each frame and the estimation errors within each frame tend to spread out evenly over the entire frame. This leads to larger relative errors at the frame boundaries, and the resulting discontinuities, which are most notable for low SNR conditions, can lead to pitch estimation errors, for example.

Discontinuities may be greatly reduced if both an analysis and synthesis windows are used in the enhancement preprocessor 8. For example, the square root of the Tukey window

$$w(i) = \begin{cases} \sqrt{0.5(1 - \cos(\pi i / M_0))} & \text{for } 1 \leq i \leq M_0 \\ \sqrt{0.5(1 - \cos(\pi(M - i) / M_0))} & \text{for } M - M_0 \leq i \leq M \\ 1 & \text{otherwise} \end{cases} \quad (1)$$

gives good performance when used as both an analysis and a synthesis window. M is the frame size in samples and M_0 is the length of overlapping sections of adjacent synthesis frames.

Windowed frames of speech data are next enhanced. This enhancement step is referenced generally as step 300 of Figure 2 and more particularly as the sequence of steps in Figures 3, 4, and 5.

2. The Transform Module

5 The windowed frames of the speech signal are output to a transform module 2, which applies a conventional fast Fourier transform (FFT) to the frame (see step 310 of Figure 3). Spectral magnitudes output by the transform module 2 are used by a noise estimation module 3 to estimate the level of noise in the frame.

10 3. The Noise Estimation Module

 The noise estimation module 3 receives as input the spectral magnitudes output by the transform module 2 and generates a noise estimate for output to the gain function module 4 (see step 320 of Figure 3). The noise estimate includes conventionally computed *a priori* and *a posteriori* SNRs. The noise
15 estimation module 3 can be realized with any conventional noise estimation technique, and may be realized in accordance with the noise estimation technique presented in the above-referenced U.S. Provisional Application No. 60/119,279, filed February 9, 1999.

4. The Gain Function Module

20 To prevent musical distortions and avoid distorting the overall spectral shape of speech sounds (and thus avoid disturbing the estimation of spectral parameters), the lower limit of the gain, G , must be set to a first value for frames which represent background noise only (a speech pause) and to a second lower value for frames which represent active speech. These limits and the gain are
25 determined illustratively as follows.

4.1 Limiting the *a priori* SNR

The gain function, G , determined by module 4 is a function of an *a priori* SNR value ξ_k and an *a posteriori* SNR value γ_k (referenced above). The *a priori* SNR value ξ_k is adaptively limited by the gain function module 4 based on whether the current frame contains speech and noise or noise only, and based on an estimated long term SNR for the speech data. If the current frame contains noise only (see step 331 of Figure 4), a preliminary lower limit $\xi_{\min 1}(\lambda) = 0.12$ is preferably set for the *a priori* SNR value ξ_k (see step 332 of Figure 4). If the current frame contains speech and noise (*i.e.*, active speech), the preliminary lower limit $\xi_{\min 1}(\lambda)$ is set to

$$\xi_{\min 1}(\lambda) = 0.12 \exp(-5)(0.5 + \text{SNR}_{\text{LT}}(\lambda))^{0.65} \quad (3)$$

where SNR_{LT} is the long term SNR for the speech data, and λ is the frame index for the current frame (see step 333 of Figure 4). However, $\xi_{\min 1}$ is limited to be no greater than 0.25 (see steps 334 and 335 of Figure 4). The long term SNR_{LT} is determined by generating the ratio of the average power of the speech signal to the average power of the noise over multiple frames and subtracting 1 from the generated ratio. Preferably, the speech signal and the noise are averaged over a number of frames that represent 1-2 seconds of the signal. If the SNR_{LT} is less than 0, the SNR_{LT} is set equal to 0.

The actual lower limit for the *a priori* SNR is determined by a first order recursive filter:

$$\xi_{\min}(\lambda) = 0.9\xi_{\min}(\lambda-1) + 0.1\xi_{\min 1}(\lambda) \quad (4)$$

This filter provides for a smooth transition between the preliminary values for speech frames and noise only frames (see step 336 of Figure 4). The smoothed lower limit $\xi_{\min}(\lambda)$ is then used as the lower limit for the *a priori* SNR value $\xi_k(\lambda)$ in the gain computation discussed below.

4.2 Determining the Gain with a Limited *a priori* SNR

As is known in the art, gain, G , used in speech enhancement preprocessors is a function of the *a priori* signal to noise ratio, ξ , and the *a posteriori* SNR value, γ . That is, $G_k = f(\xi_k(\lambda), \gamma_k(\lambda))$, where λ is the frame index and k is the subband index. In accordance with an embodiment of this invention, the lower limit of the *a priori* SNR, $\xi_{\min}(\lambda)$, is applied to the *a priori* SNR (which is determined by noise estimation module 3) the as follows:

$$\xi_k(\lambda) = \xi_k(\lambda) \text{ if } \xi_k(\lambda) > \xi_{\min}(\lambda)$$

$$\xi_k(\lambda) = \xi_{\min}(\lambda) \text{ if } \xi_k(\lambda) \leq \xi_{\min}(\lambda)$$

(see steps 510 and 520 of Figure 5).

Based on the *a posteriori* SNR estimation generated by the noise estimation module 3 and the limited *a priori* SNR discussed above, the gain function module 4 determines a gain function, G (see step 530 Figure 5). A suitable gain function for use in realizing this embodiment is a conventional Minimum Mean Square Error Log Spectral Amplitude estimator (MMSE LSA), such as the one described in Y. Ephraim et al., "Speech Enhancement Using a Minimum Mean-Square Error Log-Spectral Amplitude Estimator," IEEE Trans. Acoustics, Speech and Signal Processing, Vol. 33, pp. 443-445, April 1985, which is hereby incorporated by reference as if set forth fully herein. Further improvement can be obtained by using a multiplicatively modified MMSE LSA estimator, such as that described in D. Malah, et al., "Tracking Speech Presence Uncertainty to Improve Speech Enhancement in Non-Stationary Noise Environments," Proc. ICASSP, 1999, to account for the probability of speech presence. This reference is incorporated by reference as if set forth fully herein.

5. Applying the Gain Function

The gain, G , is applied to the noisy spectral magnitudes of the data frame output by the transform module 2. This is done in conventional fashion by multiplying the noisy spectral magnitudes by the gain, as shown in Figure 1 (see step 340 of Figure 3).

6. *The Inverse Transform Module*

A conventional inverse FFT is applied to the enhanced spectral amplitudes by the inverse transform module 5, which outputs a frame of enhanced speech to an overlap/add module 6 (see step 350 of Figure 3).

7. *Overlap Add Module; Delay Reduction*

The overlap/add module 6 synthesizes the output of the inverse transform module 5 and outputs the enhanced speech signal $\hat{s}(k)$ to the coder 7.

Preferably, the overlap/add module 6 reduces the delay imposed by the enhancement preprocessor 8 by multiplying the left "half" (e.g., the less current 180 samples) in the frame by a synthesis window and the right half (e.g., the more current 76 samples) in the frame by an inverse analysis window (see step 400 of Figure 2). The synthesis window can be different from the analysis window, but preferably is the same as the analysis window (in addition, these windows are preferably the same as the analysis window referenced in step 200 of Figure 2). The sample sizes of the left and right "halves" of the frame will vary based on the amount of data shift that occurs in the coder 7 input buffer as discussed below (see the discussion relating to step 800, below). In this case, the data in the coder 7 input buffer is shifted by 180 samples. Thus, the left half of the frame includes 180 samples. Since the analysis/synthesis windows have a high attenuation at the frame edges, multiplying the frame by the inverse analysis filter will greatly amplify estimation errors at the frame boundaries. Thus, a small delay of 2-3 ms is preferably provided so that the inverse analysis filter is not multiplied by the last 16-24 samples of the frame.

Once the frame is adjusted by the synthesis and inverse analysis windows, the frame is then provided to the input buffer (not shown) of the coder 7 (see step 500 of Figure 2). The left portion of the current frame is overlapped with the right half of the previous frame that is already loaded into the input buffer. The right portion of the current frame, however, is not overlapped with any frame or portion of a frame in the input buffer. The coder 7 then uses the

data in the input buffer, including the newly input frame and the incomplete right half data, to extract coding parameters (see step 600 of Figure 2). For example, a conventional MELP coder extracts 10 linear prediction coefficients, 2 gain factors, 1 pitch value, 5 bandpass voicing strength values, 10 Fourier magnitudes, and an aperiodic flag from data in its input buffer. However, any desired information can be extracted from the frame. Since the MELP coder does not use the latest 60 samples in the input buffer for the Linear Predictive Coefficient (LPC) analysis or computation of the first gain factor, any enhancement errors in these samples have a low impact on the overall performance of the coder 7.

After the coder 7 extracts coding parameters, the right half of the last input frame (e.g., the more current 76 samples) are multiplied by the analysis and synthesis windows (see step 700 of Figure 2). These analysis and synthesis windows are preferably the same as those referenced in step 200, above (however, they could be different, such as the square-root of the analysis window of step 200).

Next, the data in the input buffer is shifted in preparation for input of the next frame, e.g., the data is shifted by 180 samples (see step 800 of Figure 2). As discussed above, the analysis and synthesis windows can be the same as the analysis window used in the enhancement preprocessor 8, or can be different from the analysis window, e.g., the square root of the analysis window. By shifting the final part of overlap/add operations into the coder 7 input buffer, the delay of the enhancement preprocessor 8/coder 7 combination can be reduced to 2-3 milliseconds without sacrificing spectral resolution or cross talk reduction in the enhancement preprocessor 8.

C. Discussion

While the invention has been described in conjunction with specific embodiments thereof, it is evident that many alternatives, modifications and variations will be apparent to those skilled in the art. Accordingly, the preferred

embodiments of the invention as set forth herein are intended to be illustrative, not limiting. Various changes may be made without departing from the spirit and scope of the invention.

For example, while the illustrative embodiment of the present invention is presented as operating in conjunction with a conventional MELP speech coder, other speech coders can be used in conjunction with the invention:

The illustrative embodiment of the present invention employs an FFT and IFFT, however, other transforms may be used in realizing the present invention, such as a discrete Fourier transform (DFT) and inverse DFT.

While the noise estimation technique in the referenced provisional patent application is suitable for the noise estimation module 3, other algorithms may also be used such as those based on voice activity detection or a spectral minimum tracking approach, such as described in D. Malah et al., "Tracking Speech Presence Uncertainty to Improve Speech Enhancement in Non-Stationary Noise Environments," Proc. IEEE Intl. Conf. Acoustics, Speech, Signal Processing (ICASSP), 1999; or R. Martin, "Spectral Subtraction Based on Minimum Statistics," Proc. European Signal Processing Conference, vol. 1, 1994, which are hereby incorporated by reference in their entirety.

Although the preliminary lower limit $\xi_{\min 1}(\lambda) = 0.12$ is preferably set for the *a priori* SNR value ξ_k when a frame represents a speech pause (background noise only), this preliminary lower limit $\xi_{\min 1}$ could be set to other values as well.

The process of limiting the *a priori* SNR is but one possible mechanism for limiting the gain values applied to the noisy spectral magnitudes. However, other methods of limiting the gain values could be employed. It is advantageous that the lower limit of the gain values for frames representing speech activity be less than the lower limit of the gain values for frames representing background noise only. However, this advantage could be achieved other ways, such as, for example, the direct limitation of gain values (rather than the limitation of a functional antecedent of the gain, like *a priori* SNR).

Although frames output from the inverse transform module 5 of the enhancement preprocessor 8 are preferably processed as described above to reduce the delay imposed by the enhancement preprocessor 8, this delay reduction processing is not required to accomplish enhancement. Thus, the enhancement preprocessor 8 could operate to enhance the speech signal through gain limitation as illustratively discussed above (for example, by adaptively limiting the *a priori* SNR value ξ_k). Likewise, delay reduction as illustratively discussed above does not require use of the gain limitation process.

Delay in other types of data processing operations can be reduced by applying a first process on a first portion of a data frame, *i.e.*, any group of data, and applying a second process to a second portion of the data frame. The first and second processes could involve any desired processing, including enhancement processing. Next, the frame is combined with other data so that the first portion of the frame is combined with other data. Information, such as coding parameters, are extracted from the frame including the combined data. After the information is extracted, a third process is applied to the second portion of the frame in preparation for combination with data in another frame.

melp.c.

Version 1.2

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...ing the C source code software, the pre-compiled MLSP software and any
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*/
/*
/* msp.s: Mixed Excitation LPC speech coder
/*
/*

map.v: Mimed Entitation LPC speech coder

```

/* compiler include files */
#include <stdio.h>
#include <stdlib.h>
#include <math.h>
#include <unistd.h>
#include <sys/types.h>
#include <sys/stat.h>
#include <fcntl.h>
#include <sys/time.h>
#include <sys/socket.h>
#include <sys/un.h>
#include <sys/wait.h>
#include <sys/resource.h>
#include <sys/mman.h>
#include <sys/uio.h>
#include <sys/xattr.h>
#include <sys/epoll.h>
#include <sys/eventfd.h>
#include <sys/signalfd.h>
#include <sys/timerfd.h>
#include <sys/ptrace.h>
#include <sys/sem.h>
#include <sys/shm.h>
#include <sys/msg.h>
#include <sys/ipc.h>
#include <sys/types.h>
#include <sys/stat.h>
#include <sys/mman.h>
#include <sys/uio.h>
#include <sys/xattr.h>
#include <sys/epoll.h>
#include <sys/eventfd.h>
#include <sys/signalfd.h>
#include <sys/timerfd.h>
#include <sys/ptrace.h>
#include <sys/sem.h>
#include <sys/shm.h>
#include <sys/msg.h>
#include <sys/ipc.h>

```

.....
Functions added by strip


```
void mlg_unc( float speech_in[], unsigned int chn_bit(),
             struct mlg_parms *par, struct mlg_parms *new_par)
{
    unsigned int cbsuf(CMUS18);
    int i;
    int mloop;

    par->cbsuf = cbsuf;
    par->chnbit = 0;

    mlg_parms[speech_in, par, new_par];

    mloop = par->ncpsr-cbsuf;
    for (i=0; i<mloop; i++)
        chn_bit[i] = cbsuf[i];
}
```


melp_ana.c.

[illegible]

ANSI Z39.18-1997 American National Standard

Item 1.2

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Vishwanathan

Control Systems Laboratory

Dr. M.D.

Box 633103, M/S 1177

CO. TX 75265

Mixed Excitation Linear Prediction (MELP) speech coding algorithm (the "C" source code software, the pre-existing MELP software and any other software, collectively, the "Software"), and the source code of the Software, is delivered to the Government in accordance with the terms of Contract N00019-91-C-4101. It is delivered with Government license rights in the field of secure voice communications only. No other use is authorized or granted by Texas Instruments Incorporated. The Government Purpose license rights shall be effective until 30 September 2001; thereafter, the Government purpose license rights will expire and the Software shall have unlimited rights in the Software. The restrictions on the use of the Software method with this legend are set forth in the "Government Purpose license rights" in paragraph (e)(1)(i) of the contract at 352.237-701 of the contract listed above. This legend, together with the indications of the portions of this Software which are subject to Government purpose license rights shall be included on any reproduction hereof and shall include any part of the portions subject to such limitations.

Name: welp_ens.c

Description: KDP analysis

Input:

speech() - input speech signal

input

'per - KILP parameter structure

Plus d'images

mp : include /lib/

uob 40110.42

44-38861-1A

u. d. l. p. h.

under 'Approved, No.

1. 2000

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10. 11. 2019

4-214, opn

amplifier constants %

no more!

no. 0001

no impact 0.99

no POLICAR 0.99

NO PEAK - THIRTEEN

1 kmu-romu au

607-7000118

90:97:06

2

[illegible]

3

melp_ana.c.

```

20599
3:16:06
/* (outofrange == 0) {
    melp_window(speech LPC_FRAME_BD - (LPC_FRAME/2)); melp_win_coef, albuf, LPC_FRAME
    melp_autocorr(albuf, r, LPC_FRAME);
    lpc[0] = 1.0;
    melp_lpc_echu(r, lpc, refc, LPC_ORD);
}
else {
    lpc[0] = 1.0;
    melp_lpc_echu(r1, lpc, refc, LPC_ORD);
}

/*
    cout << r[0] << " " << r[1] << " " << r[2] << " " << r[3] << " " << r[4] << " " <<
    1 << " " << r[5] << " " << r[6] << " " << r[7] << " " << r[8] << " " << r[9] << " " << r[10] << "
    < endl;
    // : << r[10] << " " << r[11] << " " << r[12] << " " << r[13] << " " << r[14] << " " << r[15] << "
    < endl;
    // : << r[16] << " " << r[17] << " " << r[18] << " " << r[19] << " " << r[20] << "
    < endl;
    // cout << endl;
}

melp_lpc_bw_expand(lpc, lpc, BWFACT, LPC_ORD);

/* Calculate LPC residual */
melp_synth(speech LPC_FRAME_BD, lpc, albuf(LPC_ORD), LPC_ORD, PITCH_FR);

/* Check peakiness of residual signal */
begin = (LPC_ORD - PITCH_FR/2);
temp = melp_peakiness(albuf(begin), PITCH_FR);

/* Peakiness: force lowest band to be voiced */
if (temp > PEAK_THRESH) {
    par->bpc[0] = 1.0;
}

/* Estimate peakiness: force second and third bands to be voiced */
if (temp > PEAK_THR2) {
    par->bpc[1] = 1.0;
    par->bpc[2] = 1.0;
}

/* Calculate overall frame pitch using loopess (filtered residual) */
pitch = melp_pitch_ana(speech, pitch_frame_BD, albuf(LPC_ORD, PITCH_FR));
lpthresh = BPTHRSH;

/* Calculate gain of input speech for each gain subframe */
for (i = 0; i < MIN_GAIN_FR; i++) {
    if (par->bpc[0] > lpthresh) {
        /* voiced mode: pitch synchronous window length */
        temp = sub_pitch;
        par->gain[i] = melp_gain_ana(speech, FRAME_BD - (i * QAIN_FR),
                                temp, MIN_GAIN_FR, 2 * PITCH_FR);
    }
    else {
        temp = 1.33 * QAIN_FR - 0.5;
        par->gain[i] = melp_gain_ana(speech, FRAME_BD - (i * QAIN_FR),
                                temp, 0.2 * PITCH_FR);
    }
}

/* Update average pitch value */
if (par->gain[MIN_GAIN_FR - 1] > SILENCE_DB)
    temp = pccorr;

```

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```

track3 <- speech[limit] <- 'n', ''

/* Copy input speech to pitch window and lowpass filter */
mlp_v_eq[isigbuf(LPP_OLD),depitch[PITCH_OLD],PITCH_FR],
mlp_v_eq[isigbuf(LPP_OLD),depitch[PITCH_OLD],LPP_OLD],
mlp_polit[isigbuf(LPP_OLD),mlp_lpf_den,asigbuf(LPP_OLD),
LPP_OLD,PITCH_FR],
mlp_v_eq[isigbuf(LPP_OLD),depitch[PITCH_OLD],LPP_OLD],
mlp_serfit[asigbuf(LPP_OLD),mlp_lpf_num,asigbuf(LPP_OLD),
LPP_OLD,PITCH_FR],

/* Perform global pitch search at frame end on lowpass speech signal */
/* Note: avoid short pitches due to formant tracking */
fpitch[lim] = mlp_find_pitch[asigbuf(LPP_OLD,PITCH_FR/3),atemp,
(2*PITCHMAX),PITCHMAX],

/* Perform bandpass voicing analysis for end of frame */
mlp_bpvc[ans[ispeech[PITCH_OLD],fpitch,4par->bpvc[0],asigbuf_pitch],

/* Force jitter if lowest band voicing strength is weak */
if (par->bpvc[0] < WJTH)
    par->jitter = MAX_JITTER,
else
    par->jitter = 0.0,

/* Calculate LPC for end of frame */
if (autocorrmode == 0) {
    mlp_window[ispeech[PITCH_OLD,(LPC_FRAME/3)],mlp_win_coef,asigbuf_LPC_FRAME],
    mlp_autocorr[asigbuf_LPC_OLD,LPC_FRAME],
    lpc[0] = 1.0,
    mlp_lpc_schur(fr,lpc,refc,LPC_OLD),
}
else {
    lpc[0] = 1.0,
    mlp_lpc_schur(fr,lpc,refc,LPC_OLD),
}

/* Count <= r[0] <= ... <= r[1] <= ... <= r[2] <= ... <= r[3] <= ... <= r[6] <= ... <= r[9] <= ... <= r[10] <= end,
/* Count <= r[0] <= ... <= r[1] <= ... <= r[2] <= ... <= r[3] <= ... <= r[6] <= ... <= r[9] <= ... <= r[10] <= end,
/* Count <= r[3] <= ... <= r[6] <= ... <= r[7] <= ... <= r[10] <= ... <= r[11] <= ... <= r[12] <= ... <= r[13] <= ... <= r[14] <= ... <= r[15] <= end,
/* Count <= end,

mlp_lpc_bw_expand(lpc,lpc,IMPACT,LPC_OLD),

/* Calculate LPC residual */
mlp_serfit[ispeech[PITCH_OLD],lpc,asigbuf(LPP_OLD,LPC_OLD,PITCH_FR],

/* Check peakiness of residual signal */
begin = (LPP_OLD+(PITCHMAX/3)),
temp = mlp_peakiness[asigbuf(begin,PITCHMAX),

/* Peakiness: force lowest band to be voiced */
if (temp > PEAK_THRESH) {
    par->bpvc[0] = 1.0,
}

/* Extreme peakiness: force second and third bands to be voiced */
if (temp > PEAK_THR2) {
    par->bpvc[1] = 1.0,
    par->bpvc[2] = 1.0,
}

/* Calculate overall frame pitch using lowpass filtered residual */

```

melp_ana.c.

```

/* Quantize filter and bandpass coding */
melp_quant_vfilter(par->flt_index, 0.0, MAX_JITTER, 2);
par->uv_flag = melp_enc(par->spc[0], sps->spc_index, bpthresh,
                        MAX_BANDS);

/* Calculate Fourier coefficients of residual signal from quantised LPC */
melp_fill(par->fa_mg, 1.0, MAX_BANDS);
if (par->spc[0] > bpthresh) {
    melp_lpc_ltp(pred(par->laf, lpc, LPC_ORD));
    melp_serflt(speech(frags_end - (LPC_FRAMES/2)), lpc, algbuf,
                LPC_ORD, LPC_FRAMES);
    melp_window(buf, melp_win_coef, algbuf, LPC_FRAMES);
    melp_fltc_harm(buf, par->fa_mg, par->pitch, MAX_BANDS, LPC_FRAMES);

/* Quantize Fourier coefficients */
/* pre-weight vector, then use Euclidean distance */
melp_window(sps->fa_mg[0], v_fa_sps->fa_mg[0], MAX_BANDS);
melp_serflt(sps->fa_mg[0], sps->fa_mg[0], fa_vq_mst);

/* Set MELP indices to point to same array */
par->fa_mg_index = fa_vq_mst_index;

/* Update NSVG information */
par->nsvg_stages = vq_par.nsvg_stages;
par->nsvg_bits = vq_par.nsvg_bits;

/* Write channel bitstream */
melp_chn_write(par);

/* Update delay buffers for next frame */
if (framecode == 1) // 1st, 0/20/98
    melp_v_cmult(speech(frags_end - (LPC_FRAMES - 1), fa_vq_mst),
                melp_v_eq(speech(frags_end - (LPC_FRAMES - 1), fa_vq_mst),
                          fpitch(frags_end) - fpitch(frags_end - 1)));

/* melp_enc_init: perform initialisation */
}

void melp_enc_init(void)
{
    int j;

    melp_bvnc_ana_init(frags, PITCHWIN, PITCHWIN, MAX_BANDS, 2, MIN_HARMONICS);
    melp_pitch_ana_init(PITCHWIN, PITCHWIN, FRAMES, LTP_ORD, MIN_HARMONICS);
    melp_p_avg_init(PITCHWIN, DEFAULT_PITCH, 3);

    melp_v_spl(speech, frags, FRAMES);
    melp_v_spl(speech, lpc, frags, FRAMES);
    melp_v_spl(speech, pitch, frags, FRAMES);
    pitch_avg(DEFAULT_PITCH);
    melp_fill(fpitch, DEFAULT_PITCH, 2);
    melp_v_spl(spl, ltp, ORD);

/* Initialise multi-stage vector quantisation (read codebook) */
vq_par.nsvg_best = NSVG_M;

```

melp_ana.c.

```

/* Quantize filter and bandpass coding */
melp_quant_vfilter(par->flt_index, 0.0, MAX_JITTER, 2);
par->uv_flag = melp_enc(par->spc[0], sps->spc_index, bpthresh,
                        MAX_BANDS);

/* Calculate Fourier coefficients of residual signal from quantised LPC */
melp_fill(par->fa_mg, 1.0, MAX_BANDS);
if (par->spc[0] > bpthresh) {
    melp_lpc_ltp(pred(par->laf, lpc, LPC_ORD));
    melp_serflt(speech(frags_end - (LPC_FRAMES/2)), lpc, algbuf,
                LPC_ORD, LPC_FRAMES);
    melp_window(buf, melp_win_coef, algbuf, LPC_FRAMES);
    melp_fltc_harm(buf, par->fa_mg, par->pitch, MAX_BANDS, LPC_FRAMES);

/* Quantize Fourier coefficients */
/* pre-weight vector, then use Euclidean distance */
melp_window(sps->fa_mg[0], v_fa_sps->fa_mg[0], MAX_BANDS);
melp_serflt(sps->fa_mg[0], sps->fa_mg[0], fa_vq_mst);

/* Set MELP indices to point to same array */
par->fa_mg_index = fa_vq_mst_index;

/* Update NSVG information */
par->nsvg_stages = vq_par.nsvg_stages;
par->nsvg_bits = vq_par.nsvg_bits;

/* Write channel bitstream */
melp_chn_write(par);

/* Update delay buffers for next frame */
if (framecode == 1) // 1st, 0/20/98
    melp_v_cmult(speech(frags_end - (LPC_FRAMES - 1), fa_vq_mst),
                melp_v_eq(speech(frags_end - (LPC_FRAMES - 1), fa_vq_mst),
                          fpitch(frags_end) - fpitch(frags_end - 1)));

/* melp_enc_init: perform initialisation */
}

void melp_enc_init(void)
{
    int j;

    melp_bvnc_ana_init(frags, PITCHWIN, PITCHWIN, MAX_BANDS, 2, MIN_HARMONICS);
    melp_pitch_ana_init(PITCHWIN, PITCHWIN, FRAMES, LTP_ORD, MIN_HARMONICS);
    melp_p_avg_init(PITCHWIN, DEFAULT_PITCH, 3);

    melp_v_spl(speech, frags, FRAMES);
    melp_v_spl(speech, lpc, frags, FRAMES);
    melp_v_spl(speech, pitch, frags, FRAMES);
    pitch_avg(DEFAULT_PITCH);
    melp_fill(fpitch, DEFAULT_PITCH, 2);
    melp_v_spl(spl, ltp, ORD);

/* Initialise multi-stage vector quantisation (read codebook) */
vq_par.nsvg_best = NSVG_M;

```



melp_ana.c.

```

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vq_per.num_stages = 4;
vq_per.dimension = 10;

/*
 * Allocate memory for number of levels per stage and indices
 * and for number of bits per stage
 */
MEM_ALLOC (HALLOC, vq_per.num_levels, vq_per.num_stages, int);
MEM_ALLOC (HALLOC, vq_per.indices, vq_per.num_stages, int);
MEM_ALLOC (HALLOC, vq_per.num_bits, vq_per.num_stages, int);

vq_per.num_levels[0] = 128;
vq_per.num_levels[1] = 64;
vq_per.num_levels[2] = 64;
vq_per.num_levels[3] = 64;

v_per.num_bits[0] = 7;
v_per.num_bits[1] = 6;
v_per.num_bits[2] = 6;
v_per.num_bits[3] = 6;

vq_per.cb = melp_mvq_cb;

/* Scale codebook to 0 to 1 */
melp_v_scale(vq_per.cb, (3.0/FSAMP), 3300);

/* Initialize Fourier magnitude vector quantisation (read codebook) */
fa_vq_per.num_bits = 1;
fa_vq_per.num_stages = 1;
fa_vq_per.dimension = MIN_NAM;

/*
 * Allocate memory for number of levels per stage and indices
 * and for number of bits per stage
 */
MEM_ALLOC (HALLOC, fa_vq_per.num_levels, fa_vq_per.num_stages, int);
MEM_ALLOC (HALLOC, fa_vq_per.indices, fa_vq_per.num_stages, int);
MEM_ALLOC (HALLOC, fa_vq_per.num_bits, fa_vq_per.num_stages, int);

vq_per.num_levels[0] = FS_LEVELS;
vq_per.num_bits[0] = FS_BITS;
vq_per.cb = melp_favq_cb;

/* Initialize fixed MSB weighting and inverse of weighting */
melp_vq_few(v_fa, MIN_NAM, 60.0);

/* Pre-weight codebook (assume single stage only) */
if (melp_favq_weighted == 0)
{
    melp_favq_weighted = 1;
    for (j = 0; j < fa_vq_per.num_levels[0]; j++)
        melp_window(fa_vq_per.cb[j*MIN_NAM], v_fa,
            sfa_vq_per.cb[j*MIN_NAM], MIN_NAM);
}

```

mat_lib.c

```

for(i=0; i < n; i++)
    v1[i] = v2[i];
return(v1);
}

/* V_INNER- inner product */
float melp_v_inner(float *v1, float *v2, int n)
{
    int i;
    float innerprod;

    for(i=0; innerprod=0.0; i < n; i++)
        innerprod += v1[i] * v2[i];
    return(innerprod);
}

/* melp_v_magsq - sum of squares */
float melp_v_magsq(float *v, int n)
{
    int i;
    float magsq;

    for(i=0; magsq=0.0; i < n; i++)
        magsq += v[i] * v[i];
    return(magsq);
} /* v_magsq */

/* V_SCALE- vector scale */
float melp_v_scale(float *v, float scale, int n)
{
    int i;

    for(i=0; i < n; i++)
        v[i] *= scale;
    return(v);
}

/* V_MULT - componentwise vector multiplication, RM 07/20/98 */
float melp_v_mult(float *v1, float *v2, int n)
{
    int i;

    for(i=0; i < n; i++)
        v1[i] *= v2[i];
    return(v1);
}

/* V_DIV - componentwise vector division, RM 07/20/98 */
float melp_v_div(float *v1, float *v2, int n)
{
    int i;

    for(i=0; i < n; i++)
        v1[i] /= v2[i];
    return(v1);
}

/* V_SUB- vector difference */
float melp_v_sub(float *v1, float *v2, int n)
{
    int i;

    for(i=0; i < n; i++)

```

4 type MSLP Federal Standard speech coder

return 1.2

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mat_lib.c: Matrix and vector manipulation library

#include "spbald.h"

#include "mat.h"

/* vector addition */

void v_add(float *v1, float *v2, int n)

{

int i;

for(i=0; i < n; i++)

v1[i] += v2[i];

return(v1);

/* vector equate */

void v_eq(float *v1, float *v2, int n)

{

int i;

for(i=0; i < n; i++)

v1[i] = v2[i];

return(v1);

/* melp_v_eqv_int(int *v1, int *v2, int n)

{

int i;

2

mat_lib.c.

```

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v[i] = v2[i];
return(v);

/* melp_v_vap - clear vector */
float *melp_v_vap(float *v, int n)
{
    int i;
    for(i=0; i < n; i++)
        v[i] = 0.0;
    return(v);
    /* V_VAP */
}

/* melp_v_vap_int(int *v, int n)
{
    i;
    for(i=0; i < n; i++)
        v[i] = 0;
    return(v);
    /* V_VAP */
}

```


dsp_sub.c.

```

/* Subroutine envelope: calculate time envelope of signal. */
/* Note: the delay history requires one previous sample */
/* of the input signal and two previous output samples. */
/*
/* Define C2 (-0.9409)
/* Define C1 1.9266
void melp_envelope(float input[], float prev_in, float output[], int npts)
{
    int i;
    float curr_abc, prev_abc;
    prev_abc = fabs(prev_in);
    for (i = 0; i < npts; i++) {
        curr_abc = fabs(input[i]);
        output[i] = curr_abc - prev_abc + C2*output[i-2] + C1*output[i-1];
        prev_abc = curr_abc;
    }
}

/* Subroutine fill: fill an input array with a value.
/*
/* void melp_fill(float output[], float fillval, int npts)
/*
/* int i;
/* for (i = 0; i < npts; i++)
/*     output[i] = fillval;
/*
/* Subroutine interp_array: interpolate array
/*
/* void melp_interp_array(float prev[], float curr[], float out[],
/*     float ifact, int size)
/*
/* int i;
/* float ifact2;
/* ifact2 = 1.0 - ifact;
/* for (i = 0; i < size; i++)
/*     out[i] = ifact*curr[i] + ifact2*prev[i];
/*
/* Subroutine median: calculate median value
/*
/* #define MAXSORT 5
/* float melp_median(float input[], int npts)
/*
/* int i, j, loc;
/* float insert_val;
/* float sorted[MAXSORT];
/* sort data in temporary array */
/*
/* #define PRINT
/* if (npts > MAXSORT) {
/*     printf("ERROR: median else too large.\n");
/*     exit(1);
/* }

```

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4 kbps MELP Federal Standard speech coder

version 1.2

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dsp_sub.c: general subroutines.

```

/* compiler include files */
#include <stdio.h>
#include <stdlib.h>
#include <math.h>
#include "dsp_sub.h"
#include "spbd.h"
#include "mat.h"

#define SHORT_SPEECH
#define PRINT 1

/* Subroutine autocorr: calculate autocorrelations
/*
/* void melp_autocorr(float input[], float r[], int order, int npts)
/*
/* int i;
/* for (i = 0; i < order; i++)
/*     r[i] = melp_v_inner(input[i], input[i], (npts-i));
/* if (i[0] < 1.0)
/*     r[0] = 1.0;

```

dsp_sub.c.

```

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void
melp_v_eqsort(int npts)
{
    for (i = 0; i < npts; i++) {
        /* for each data point */
        for (j = 0; j < i; j++) {
            /* find location in current sorted list */
            if (sorted[j] < sorted[i])
                break;
        }
        /* Insert new value */
        loc = j;
        insert_val = sorted[i];
        for (j = i; j > loc; j--)
            sorted[j] = sorted[j-1];
        sorted[loc] = insert_val;
    }
    return(sorted(npts/2));
}

/* Subroutine PACK_CODES, Pack bit code into channel. */
/*
id melp_pack_code(int code, unsigned int *p_ch_beg, int *p_ch_bit,
int numbits, int nvals)
{
    int i, ch_bit;
    unsigned int *ch_word;

    ch_bit = *p_ch_bit;
    ch_word = *p_ch_beg;

    for (i = 0; i < numbits; i++) {
        /* Pack in bit from code to channel word */
        if (ch_bit == 0)
            *ch_word = (i < 1) ? 1 : 0;
        else
            *ch_word |= ((i < 1) ? 1 : 0) << ch_bit;

        /* Check for full channel word
        if (i < ch_bit && wait) {
            ch_bit = 0;
            (*p_ch_beg)++;
            ch_word++;
        }
    }

    /* Save updated bit counter */
    *p_ch_bit = ch_bit;

    /* Subroutine peakness, estimate peakness of input
    aligned using ratio of L3 to L1 norms.
    */
    /*
    void melp_peakness(float input[], int npts)

```

dsp_sub.c.

```

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dsp_sub.c.
/* Quantize input to correct level */
/* p_in = gain * (i * step);
/* p_index = i;

Subroutine QUANT_U_DEC: decode uniformly quantized
value.
id melp_quant_u_dec(int index, float *p_data, float gain, float qmax,
int nlen)
{
    float float_step;

    /* Define symmetrical quantizer stepsize
    step = (qmax - gain) / (nlen - 1);

    /* Decode quantized level
    *p_data = gain + (index * step);

Subroutine rand_num: generate random numbers to fill
array using system random number generator.
id melp_rand_num(float output[], float amplitude, int npts)
{
    int i;
    for (i = 0; i < npts; i++) {
        /* use system random number generator from -1 to +1 */
        /* RAND_MAX environment */
        output[i] = (amplitude*2.0) * ((float) rand()/(RAND_MAX + 0.5));
    }
    /* Assume Sun OS4 */
    output[i] = amplitude * (float) ((random() >> 16)/32767. - .5)*2);
}

Subroutine READBL: read block of input data
define MAXSIZE 1024
id melp_readbl(float input[], FILE *fp_in, int else)
{
    int i, length;
    SPBCTN int_sp[MAXSIZE]; /* Integer input array

iddef PRINT
    if (false > MAXSIZE) {
        /* Check for end of channel word
        exit();
    }
    length = fread(int_sp, sizeof(int), else, fp_in);
}

```

```

/* Subroutine serfilt, all zero (VIN) filter.
/* Notes: the output array can overlay the input.
/*
void melp_serfilt(float input[], float coeff[], float output[],
int order, int npts)
{
    int i, j;
    float accum;

    for (i = npts-1; i >= 0; i--) {
        accum = 0.0;
        for (j = 0; j <= order; j++)
            accum += input[i-j] * coeff[j];
        output[i] = accum;
    }
}

```

27.

main.c

```

float r1(LPC_ORD_t);

char in_name[80], in_name_lpc[80], in_name_pitch[80], out_name[80], out_params_name[80],
in_params_name[80], autocorr_name[80];

void main(int argc, char **argv)
{
    void parse(int argc, char **argv);

    int inFrame;
    float *speech_in, *speech_in_lpc, *speech_in_pitch;

    int length, frame, eof_reached;
    int num_frames = 0;

    // RM, 07/20/98
    // float speech_inFrame;
    // float speech_outFrame;
    // static struct melp_param melp_par;
    // static struct melp_param new_par;
    // /* new melp parameters */

    unsigned int chan_bit(CHSIZE);
    FILE *fp_in, *fp_out, *fp_in_lpc, *fp_in_pitch;
    FILE *fp_autocorr;

    /* Print user message */
    printf("\n4.4 kHz Federal Standard NLP speech coder\n");
    printf("C simulation, version RM 1a1a\n");

    //***** Get input parameters from command line *****
    parse(argc, argv);

    ofstream track(out_params_name);
    if (!track) { if (writeMode == 1) { cerr << "Cannot open " << out_params_name <<
        " output file!" << endl;
        exit(1);
    }
    }

    if (input file) << endl;
    exit(1);

    if (inFrame < readMode == 1) { cerr << "Cannot open " << in_params_name <<
        " input file!" << endl;
        exit(1);
    }

    if (autocorrMode == 1) {
        if (!fp_autocorr) { fp_autocorr = fopen(autocorr_name, "rb"); if (NULL) {
            printf("ERROR: cannot read file %s.\n", autocorr_name);
            exit(1);
        }
    }

    if (frameMode == 1) {
        inFrame = inFrame;
        NEW_ALLOC (inFrame, speech_in, inFrame, float); // RM, 07/20/98
        NEW_ALLOC (inFrame, speech_in_lpc, inFrame, float);
        NEW_ALLOC (inFrame, speech_in_pitch, inFrame, float);
    }
    else {
        inFrame = FRAME;
    }
}

```

02/05/99
16:17:09

```

/* kHz NLP Federal Standard speech coder
version 1.2

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Shu Vlasovathan
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Corporate R&D
Texas Instruments
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Dallas, TX 75265

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/*
* melp.c: Mixed Excitation LPC speech coder
*
*
* compiler include files */
#include <stdio.h>
#include "melp.h"
#include "speech.h"
#include "melp.h"
#include <fstream.h>

/* compiler constants */
#define AMA_BW 0
#define ANALYSIS 1
#define SYNTHESIS 2

/* note: CHSIZE is shortest integer number of words in channel packet */
#define CHSIZE 9
#define NUM_CH_BITS 50

/* external memory */
int readMode = AMA_BW;
int frameMode = 0;
int autocorrMode = 0;
int inFrame = 0;
int writeMode = 0;
int readMode = 0;
int frameMode = 0;
int pitchMode = 0;

```

```

// RM, 07/20/98
MEM_ALLOC(INALLOC, speech_in_frames, float);
MEM_ALLOC(INALLOC, speech_in_lpc, FRAME, float);
MEM_ALLOC(INALLOC, speech_out_pitch, FRAME, float);
},

/* Open input, output, and parameter files */
if (!fp_in = fopen(in_name, "rb")) == NULL || {
    printf(" ERROR: cannot read file %s.\n", in_name);
    exit(1);
}

if (lpcmode == 1) {
    if (!fp_in_lpc = fopen(in_name_lpc, "rb")) == NULL || {
        printf(" ERROR: cannot read lpc file %s.\n", in_name_lpc);
        exit(1);
    }
}

if (pitchmode == 1) {
    if (!fp_out_pitch = fopen(out_name_pitch, "wb")) == NULL || {
        printf(" ERROR: cannot write file %s.\n", out_name_pitch);
        exit(1);
    }
}

/* Check length of channel input if needed */
if (lpcmode == SYNTHESIS) {
    fseek(fp_in, 0L, 2);
    length = ftell(fp_in);
    rewind(fp_in);

    num_frames = 0.5 * length * (8.0 / WORD_CN_BITS) * (6.0/32);

    int input_num, length_lpc, length_pitch;

    /* Check length of input file if needed, RM, 07/20/98 */
    if (lpcmode != SYNTHESIS && (framemode==1)) {
        fseek(fp_in, 0L, 2);
        length = ftell(fp_in);
        rewind(fp_in);

        if (lpcmode == 1) {
            if ((length/sizeof(float)) % inframe != 0) {
                printf(" ERROR: file %s must contain a multiple of %d samples.", in_name, inframe);
                exit(1);
            }
        } else {
            if ((length/sizeof(short)) % inframe != 0) {
                printf(" ERROR: file %s must contain a multiple of %d samples.", in_name, inframe);
                exit(1);
            }
        }
    }

    input_num = length/sizeof(float) / inframe;

    if (lpcmode == 1) {
        if ((length/sizeof(float)) % inframe != 0) {
            printf(" ERROR: file %s must contain a multiple of %d samples.", in_name, inframe);
            exit(1);
        }
    } else {
        if ((length/sizeof(short)) % inframe != 0) {
            printf(" ERROR: file %s must contain a multiple of %d samples.", in_name, inframe);
            exit(1);
        }
    }

    /* Initialize MLP analysis and synthesis */
    if (lpcmode != SYNTHESIS)
        mlp_enc_init();
}

```


02/05/99
16:17:02
--WV120004

main.c.

```

main:
    melp_mode = ANALYSIS;
    while (!--argc) && (!(*argv)[0] == '-'){
        switch ((*argv)[1]){
            case 'a':
                melp_mode=ANALYSIS;
                break;
            case 's':
                melp_mode=SYNTHESIS;
                break;
            case 't':
                ascanf((*argv,"%s",in_name);
                --argc;
                break;
            case 'o':
                ascanf((*argv,"%s",out_name);
                --argc;
                break;
            case 'e':
                // RM, 07/30/98 read input signal
                in_frames = 256;
                break;
            case 'r':
                // RM, 08/06/98 read additional f
                autocorrmode = 1;
                ascanf((*argv,"%s",autocorr_name);
                --argc;
                break;
            case 'f':
                // RM, 08/06/98 read input file f
                floatmode = 1;
                break;
            case 'w':
                // RM, 08/06/98 write parameters
                writemode = 1;
                ascanf((*argv,"%s",out_params_name);
                --argc;
                break;
            case 'c':
                // RM, 08/06/98 write parameters
                readmode = 1;
                ascanf((*argv,"%s",in_params_name);
                --argc;
                break;
            case 'l':
                // RM, 08/24/98 read separate fil
                lpcm_mode = 1;
                ascanf((*argv,"%s",in_name_lpcm);
                --argc;
                break;
            case 'p':
                // RM, 08/24/98 read separate fil
                pitchmode = 1;
                ascanf((*argv,"%s",in_name_pitch);
                --argc;
                break;
            case 'h':
                // RM, 08/06/98 apply input filter
                filtermode = 1;
                break;
            default:
                break;
        }
    }

    if (lpcm_mode == 1) && (pitchmode == 1)
        melp_out(speech_in, speech_in_lpcm, speech_in_pitch, chan_
    )
    else {
        // invalid mode
        printf(" error, invalid read/write mode! /n");
        exit(1);
    }

    /* If ANALYSIS mode only */
    if (melp_mode == ANALYSIS && melp_per_chbit == 0) {
        /* write channel output to file */
        fwrite (invald, chan_bit, sizeof (int),
                CHSIS, fp_out);
    }

    /* Perform MZLP synthesis (skip first frame) */
    if (melp_mode != ANALYSIS) {
        /* If SYNTHESIS mode only */
        if (melp_mode == SYNTHESIS && melp_per_chbit == 0) {
            /* Read channel input from file */
            fread((void *) chan_bit, sizeof (int), CHSIS,
                  fp_in);
        }

        melp_dec(chan_bit, speech_out, &melp_per);

        if (frame > 0)
            melp_writeb(speech_out,fp_out,FRAME);
    }

    frame++;
    if (melp_mode == SYNTHESIS) {
        if (frame == num_frames)
            eof_reached = 1;
    }

    fclose(fp_in);
    fclose(fp_out);

    if (!tracbl, bad()) & writemode == 1)
        cerr << "file " << out_params_name << " had some problems." << endl;

    if parse(int argc, char **argv) {
        int error_flag;
        error_flag = 0;
        if (argc < 2)
            error_flag = 1;
    }

```




main.c.

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-0500

```

error_flag = 1;
break;
}

if (error_flag == 1) {
    printf(stderr, "usage:\n");
    printf(stderr, "analysis/synthesis: melp -i infile -o outfile\n");
    printf(stderr, "analysis only: melp -e -i infile -o outfile\n");
    printf(stderr, "synthesis only: melp -s -i bitfile -o outfile\n");
    printf(stderr, "\n");
    printf(stderr, "-i : speech input file in float format\n");
    printf(stderr, "-e : speech input file contains frames of 136 samples each\n");
    printf(stderr, "-s : read additional autocorrelation values in\n");
    printf(stderr, "float format from file <filename>.\n");
    printf(stderr, "-w <filename> : Write MELP parameters to file <filename>.\n");
    printf(stderr, "-c <filename> : Read MELP parameters from file <filename>.\n");
    printf(stderr, "-w option\n");
    printf(stderr, "-i <filename> : Read lpc info from file <filename>.\n");
    printf(stderr, "-p <filename> : Read pitch info from file <filename>.\n");
    printf(stderr, "-h : Apply input highpass filter to input files.\n");
    exit(1);
}

if (melpmode == ANALYSIS)
    printf("MELP analysis and synthesis\n");
else
    if (melpmode == ANALYSIS)
        printf("MELP analysis\n");
    else
        if (melpmode == SYNTHESIS)
            printf("MELP synthesis\n");
    if (framesize == 1)
        printf("... Read enhanced frames of size 136000.\n");
    printf("Input from stdin output to %s.\n", in_name, out_name);

```



dsp_sub.h.

```

void melp_xand_num(float output[], float amplitude, int npts);
int melp_unpack_code(unsigned int *p_ch_beg, int *p_ch_bit, int *p_code,
    int numbits, int erase, unsigned int erase_mask);
void melp_window(float input[], float win_coef[], float output[], int npts);
void melp_srfilt(float input[], float coeff[], float output[], int order, int npts);
int melp_readbl(float input[], float *p_in, int nsize);
int melp_readbl_float(float input[], float *p_in, int nsize); /* RM */
void melp_writebl(float output[], float *p_out, int nsize);

```

02/05/99
16:17:34

4 type NSLP Federal Standard speech coder
version 1.2

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John Vlasovathan

Personal Systems Laboratory

Corporate R&D

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dsp_sub.h: include file

```

#ifndef __FLOAT__
#define double float;
#define float

```

/* External function definitions */

```
void melp_autocorr(float input[], float r[], int order, int npts);
```

```
void melp_envelope(float input[], float prev_in, float output[], int npts);
```

```
void melp_filt(float output[], float filter, int npts);
```

```
void melp_interp_array(float prev[], float curr[], float out[], float start, int nsize);
```

```
float melp_median(float input[], int npts);
```

```
void melp_pack_code(int code, unsigned int *p_ch_beg, int *p_ch_bit, int numbits,
    int nsize);
```

```
float melp_peakiness(float input[], int npts);
```

```
void melp_polfilt(float input[], float coeff[], float output[], int order, int npts);
```

```
void melp_quant_u(float *p_data, int *p_index, float gain, float qmax, int nsize);
```

```
void melp_quant_u_dec(int index, float *p_data, float gain, float qmax, int nsize);
```

mat.h.

16:18:05
 16:18:05
 16:18:05

1 kbps MELP Federal Standard speech coder

version 1.2

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Shu Vireenathan

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mat.h Matrix include file.
 (Low level matrix and vector functions.)

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```

#define _FLOAT_
#include <math.h>
typedef float _FLOAT_;

void mat_v_invert(float *v1, float *v2, int n);
void mat_v_multiply(float *v1, float *v2, int n);
void mat_v_scale(float *v1, float *v2, int n);
void mat_v_add(float *v1, float *v2, int n);
void mat_v_sub(float *v1, float *v2, int n);
void mat_v_eq(float *v1, float *v2, int n);
void mat_v_neq(float *v1, float *v2, int n);
void mat_v_div(float *v1, float *v2, int n);
void mat_v_mult(float *v1, float *v2, int n);

```

02/05/99
15:58:36

enhance.c.

```

-----
enhance.c - Main program for MSE-LSA speech enhancement with
             minimum statistics noise estimation
-----
* Authors: Rainer Martin, AT&T Labs-Research
* Last Update: 8/18/8
-----
#include "enhance.h"
#include "vect_fun.h"
#include "enh_fun.h"

static char in_name[80], out_name[80], version_name[40];

int i;

int i;
int i;

Enhance_Param P;
Enhance_Delta D;

FILE *fp_in, *fp_out;
float *speech_in_float, *speech_out_float, *speech_overlap_float, *new_speech;
short *speech_short, *noises;

/* Get input parameters from command line and overwrite default parameters if necessary */
parse_command_line(argc, argv);
init_params(&P, &D, version_name); /* Initialize parameters with default values */

speech_short = CALLOC_SHORT(P.win_shift); /* buffer for input samples */
speech_in_float = CALLOC_FLOAT(P.win_len); /* buffer of input samples of one frame */
speech_out_float = CALLOC_FLOAT(P.win_len); /* buffer of input samples of one frame */
speech_overlap_float = CALLOC_FLOAT(P.win_len); /* buffer of input samples of one frame */
new_speech = &speech_in_float[P.overlap_len]; /* pointer on new data */

/* Print user message */
printf("MSE-LSA speech enhancement with Minimum Statistics noise estimation.\n");
printf("C simulation, version 1.0\n");
printf("Input from: %s\n", in_name);
printf("Output to: %s\n", out_name);

/* Open input, output, and parameter files */
if (fp_in = fopen(in_name, "r")) {
    if (fp_out = fopen(out_name, "w")) {
        printf("ERROR: cannot write file %s. (enhance)\n", out_name);
        exit(1);
    }
}

/* peek at first frames for noise estimate */
n_initial_noise = CALLOC_FLOAT(P.noise_frames * P.win_shift * P.win_shift);
fread(n_initial_noise, sizeof(float), P.noise_frames * P.win_shift * P.win_shift, fp_in);

```


02/05/99
15:59:08

```

#include "enhance.h"
#define _enhance_

"enhance.h - Data Structures For All Enhancement Routines
Author: Reiner Martin, AT&T Labs-Research
Last Update: 8/10/8

.....

"include "globals.h"
"include "treat.h"

..... PARAMETERS .....
" struct noise_param_Malah {
    int wtr_front_len;
    float hear_thr_rms;
    float env_rater;
    float alpha_env;
    float beta_env;
    float rean_thr;
    float CN_MIN;

    float f_n_min; /* Min. Noise-update factor - Noise only */
    float f_n_max; /* Max. Noise-update factor - Noise only */
    float f_n_min; /* Min. Noise-update factor - Signal present */
    float f_n_max; /* Max. Noise-update factor - Signal present */
    float nsmh_bias; /* bias factor for initial noise estimate */
    float noise_b_by_nsmh_b;
    float GAMMA_TH; /* Gamma_av upper threshold */
    float gamma_max_thr; /* Gamma_max threshold */
    float gamma_av_thr; /* Gamma_av threshold */
    float gamma_max_thr; /* Gamma threshold - signal present */

    No..._Param_Malah;

    typedef struct noise_param_minstat {
        int wtr_front_len;
        float rean_thr;
        float hear_thr_rms;
        float CN_MIN;

        float nsmh_bias; /* bias factor for initial noise estimate */
        float noise_b_by_nsmh_b;
        float GAMMA_TH; /* Gamma_av upper threshold */
        float gamma_max_thr; /* Gamma_max threshold */
        float gamma_av_thr; /* Gamma_av threshold */
        float gamma_max_thr; /* Gamma threshold */

        int len_minwin; /* length of subwindow */
        int num_minwin; /* number of subwindows */
        int CN_min; /* Total length of window for minimum search */
    } Param_Minstat;

```

enhance.h

```

    float alpha_k_max; /* maximum of time varying smoothing parameter */
    float Pdecy_num;

    float minv;
    float minv_aub;
    float Pvar;
    float Pvar_aub;
} Noise_Param_Minstat;

typedef struct enhance_param {
    int ZWLP_FLG; /* Set to 0 for no lower envelope tracking
                  Set to 1 for tracking with Spectral Sampling
                  Set to 2 for tracking with Mean Matching only */
    int NS_FLG; /* Set to 1 to update noise_spec - signal present */
    int fs;
    int win_len;
    float win_len_inv;
    int win_shift;
    int overlap_len;
    float win_ratio;
    float win_shift_ratio;
    int vec_len;
    float rate;
    float rate_factor;
    int noise_frames;
    float noise_bias;
    float qh;
    float alpha_LF;
    float beta_LF;
    float qh_max; /* Max value for prob. of signal absence */
    float qh_min; /* Min value for prob. of signal absence */
    float alpha_k; /* decision directed ksl weight */
    float beta_k;
    float alpha; /* 11 recursive (inter_frame) smth factor */
    float beta;
    float ksl_min;
    float vb;
    float gamma_q_thr;
    float alpha_q; /* 'false alarm' prob. for setting thld to find qh's */
    float beta_q; /* Smoothing factor of hard-decision qh-value */
    int software_ver;
    float analysis_window;

    typedef MALAH
    Noise_Param_Malah *MP;
    Noise_Param_Minstat *MP;
} Enhance_Param;

```



enhance.h.

02/05/99
15:59:08

```

Enhancement state information */
#define struct enhance_data {
    int i;
    float *initial_noise;
    float *qt;
    float *qlr;
    float *lambdab;
    float *y10;
    float *Aest;
    float *temp1;
    float *temp2;
    float *temp3;
    float *temp4;
    float *amely;
    float *y;
    float *y1;
    float *yest;
    float *gamat;
    float *h;
    float *hsl;
    float *Gin;
    float *G0;
    float *GinD;
    float *v0;
    float *yest;
    float *M_jwr;
    float *CM_min;
    float *Ksl_min;
    float *Ksl_min_var;

    float *Y_LT;
    float *SN_LT;
    float *SN_LT0;

    float *cache;

    float *M_jwr;
    float *ndiff;

    /* only for Melah's noise estimation */
    float *Y1_sach;
    int *c10;
    int *c01;
    float *Y1;
    float *Y10_av;
    float *Y1_av;
    float *M_jwr0;

    float *env10;
    int *env_flg;
    int *env_drop_flg;
    int *updn_flg;
    int *Y1_flg;
    int *Y1;

    /* only for minimum statistics */
    float *smoothed_spect;
    float *biased_smoothed_spect;
    float *biased_smoothed_spect_sub;
    float *c10_min;
    float *act_min;
    float *act_min_sub;
    float *noisespect;
    float *alpha_var;
};

/* ring buffer */
/* ring buffer pointer */

float **c10rb;
int c10rb_ind;

short *local_lag;
int minspec_counter;
float alphacorr;

float *var_ap_av;
float *var_ap_3;
float *var_rel;
float var_rel_av;

#endif

) Enhance_Data;

void parse_command_line(int argc, char **argv);

#endif

```

11/2/05/99
15:59:28

```
#include "globals.h"
#include "enh_fun.h"
#include "enhance.h"
#include "vect_fun.h"
#include "windows.h"
#include "string.h"
#include "float.h"
```

```
/* ..... Speech Enhancement Functions
Author: Rainer Martin, AT&T Labs Research
Last Update: $Id$
.....
```

```
.....
Subroutine CALLOC_FLOAT: memory allocation for float vectors
.....
it CALLOC_FLOAT(int num_samples)
```

```
float* tmp;
mp = calloc(num_samples,sizeof(float));
if (tmp == NULL)
```

```
print("ERROR: CALLOC_FLOAT request cannot be satisfied! \n\n");
terminate();
}
```

```
return tmp;
```

```
.....
Subroutine CALLOC_FLOATP: memory allocation for pointers to
float vectors
.....
it CALLOC_FLOATP(int num_samples)
```

```
float** tmp;
mp = calloc(num_samples,sizeof(float*));
if (tmp == NULL)
```

```
print("ERROR: CALLOC_FLOATP request cannot be satisfied! \n\n");
terminate();
}
```

```
return tmp;
```

```
.....
Subroutine CALLOC_SHORT: memory allocation for short vectors
.....
it CALLOC_SHORT(int num_samples)
```

enh_fun.c

```
{
short* tmp;
tmp = calloc(num_samples,sizeof(short));
if (tmp == NULL)
{
print("ERROR: CALLOC_SHORT request cannot be satisfied! \n\n");
terminate();
}
return tmp;
}
.....
/* Subroutine terminate: terminate enhancement program
.....
void terminate(int error_no)
{
printf("Program exit with error code: %d\n",error_no);
exit();
}
.....
typedef MALAM
.....
/* Subroutine init_noise_params_malah: initialise parameters of noise
/* estimation procedure
.....
void init_noise_params_malah(Enhance_Par* p)
{
p->NP->wtr_front_len = 32;
p->NP->hear_thr_rms = 6.0; /* hearing threshold rms */
p->NP->env_rate = 1. + 0.02 * p->win_ratio; /* lower envelope rise in dB */
p->NP->sigpho_env = 1. - 1e-6 * p->win_ratio * p->win_shift_ratio; /* lower envelope
parameter */
p->NP->beta_env = 1. - p->NP->alpha_env;
typedef USDOUBLES
p->NP->resn_thr = 20 * log10 (p->NP->hear_thr_rms); /* desired residual sbe noise is
vel */
else
p->NP->resn_thr = 20 * log10 (p->NP->hear_thr_rms); /* desired residual sbe noise is
vel */
endif
p->NP->CH_MIN = 0.12; /* max value for Min. Gain Modif Factor - CH_min */
p->NP->f_p_min = 0.01 * p->win_ratio * p->win_shift_ratio; /* Min. Noise-update fact
or - Noise only */
p->NP->f_p_max = 0.1 * p->win_ratio * p->win_shift_ratio; /* Max. Noise-update fact
or - Noise only */
p->NP->f_p_min = 0.02 * p->win_ratio * p->win_shift_ratio; /* Min. Noise-update fact
or - Signal present */
p->NP->f_p_max = 0.10 * p->win_ratio * p->win_shift_ratio; /* Max. Noise-update fact
or - Signal present */
p->NP->smath_bise = 1 + (1 - p->win_ratio) / 3.0; /* bias factor for initial noi
se estimate */
p->NP->noise_b_by_smath_b = p->noise_bise * p->NP->smath_bise;
typedef USDOUBLES
p->NP->gamma_TH = 2. * log ( 3. ); /* Upper third on gamma_ev for noise on
ly cond. */
else
p->NP->gamma_TH = 3. * log ( 3. );
}
```


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```

p->win_len = 256;
p->win_len_inv = 1.0 / p->win_len;

if (p->software_var == 7) {
    p->win_shift = 180; /* frame advance suitable for MELP coder */
    p->analysis_window = sqrt_tukay;
}

else {
    p->win_shift = 128; /* standard frame advance */
    p->analysis_window = hann_even;
}

p->overlap_len = p->win_len - p->win_shift;
p->win_ratio = p->win_len / 256;
p->win_shift_ratio = 2.0 * p->win_shift / p->win_len;

p->lenf = p->win_len / 2 + 1;
p->rate = 8;
p->rate_factor = p->rate;

#ifdef USEDOUBLES
    /* noise spectral bias factor (1.5dB) */
    p->noise_bias = sqrt(12.0);
    /* noise spectral bias factor (1.5dB) */
    p->noise_bias = sqrt(12.0);
    /* noise spectral bias factor (1.5dB) */
    p->qm = 1 / p->noise_bias;
    p->noise_frames = 1;

    p->alpha_LF = 1.0 - (1.0 / (float)((1.0 * p->win_shift) * 1.0)) * p->win_ratio;
    p->beta_LF = 1.0 - p->alpha_LF;

    p->qh_max = 1. - 0.001;
    p->qh_min = 0.001;

    if (p->software_var == 6)
        p->alphah = 0.99 - (p->rate / 16.0) * 0.08;
    else
        p->alphah = 0.99 - (p->rate / 16.0) * 0.12;

    p->wsh = 1. - p->alphah;
    p->phay = 1. - p->win_ratio;
    p->betay = 1. - p->alphay;

    p->shf_minq = 0.2;

#ifdef USEDOUBLES
    p->vb = log(1. / (1. - 0.5)); /* 'false alarm' prob. for setting thld to find qh's
    */
    p->vb = log(1. / (1. - 0.5)); /* 'false alarm' prob. for setting thld to find qh's
    */
    p->gammag_thr = (1. - p->shf_minq) * p->vb * p->qh;

    /* Smoothing factor of hard decision qh value */
    p->alphag = 0.99 * p->win_ratio * 0.01 * p->win_shift_ratio;
    p->betaeq = 1. - p->alphag;

    /*if MALAH
    p->NP = malloc(sizeof(Noise_Params_Malah));
    init_noise_params_malah(p);
    else
    p->NP = malloc(sizeof(Noise_Params_Minstat));
    */

```

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```

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.....
/* Subroutine compute_qt_new: compute the probability of speech absence */
/* This uses an harddecision approach due to Malah (ICASSP 1993). */
.....
loat "compute_qt_new(Float *qh,Float *qla,Float *gamet,Float *gamet0,TM,Float alpha
Float betaq,Int m)
{
    Int i;
    for ( i = 0; i < m; i++)
    {
        qla[i] = alpha*q[qla[i]];
        if (gamet[i] < Gamet0_TM)
            qla[i] += betaq;
        qh[i] = qla[i];
    }
    return(qh);
}

.....
/* Subroutine gain_log_mse: compute the gain factor and the auxiliary */
/* variable vk for the Ephraim-Malah 1985 log spectral estimator. */
/* Approximation of the exponential integral due to Malah, 1996 */
.....
loat "gain_log_mse(Float *Gain,Float *vk,Float *qh,Float *kai,Float *gamet,Int m)
{
    Int i;
    Float temp, kalg_inv, kai_vq, elv;
    for ( i = 0; i < m; i++)
    {
        temp = 1. - qh[i];
        kalg_inv = temp / kai[i];
        kai_vq = 1. / ( 1. + kalg_inv );
        vk[i] = kai_vq * gamet[i];
        if (vk[i] < 2.22044609250313e-16)
            vk[i] = 2.22044609250313e-16; /* MATLAB eps */
    }
    ..(def USEDOUBLES
        if (vk[i] < 0.1)
            elv = -2.31 + log10(vk[i]) - 0.6;
        else
            if (vk[i] > 1)
                elv = pow(10, -0.52 + vk[i] - 0.26 );
            else
                elv = -1.544 + log10(vk[i]) + 0.166;
        Gain[i] = kai_vq * exp (0.5 * elv);
    ..else
        if (vk[i] < 0.1)
            elv = -2.31 + log10(vk[i]) - 0.6;
        else
            if (vk[i] > 1)
                elv = pow(10, -0.52 + vk[i] - 0.26 );
            else
                elv = -1.544 + log10(vk[i]) + 0.166;
        Gain[i] = kai_vq * exp( 0.5 * elv);
    }
}

.....
/* Subroutine kst_min_adapt: computes the adaptive kst_min */
/* kst_min_adapt(int n_flag,Float kst_min,Float en_ll)
{
    Float kst_min_new;
    if (n_flag == 1) /* AM: adaptive modification of kst limit (10/98) */
        kst_min_new = kst_min;
    else
    {
        kst_min_new = kst_min*exp(0.65*log10.5*en_ll) - 5);
        kst_min_new = kst_min*exp(0.65*log10.5*en_ll) - 5);
    }
    if (kst_min_new > 0.25)
        kst_min_new = 0.25;
    return(kst_min_new);
}

.....
/* Subroutine smoothing_win: applies the Pارسن window */
/* void smoothing_win ( Float *tempv2, Enhance_Param *P) {
    Int i;
    Float *fp;
    fp = tempv2 + P->win_len - 1;
    for ( i = 1; i < P->HP->win_front_len; i++, fp++) {
        tempv2[i] += *fp;
        fp[i] += *fp;
    }
    while ( i < P->win_len - P->HP->win_front_len + 1 )
        tempv2[i++] = 0.0;
}

.....
/* Subroutine gain_log_mse: compute the gain factor and the auxiliary */
/* variable vk for the Ephraim-Malah 1985 log spectral estimator. */
/* Approximation of the exponential integral due to Malah, 1996 */
.....
void track_envelope(Float *Vt_av, Enhance_Data *D, Enhance_Param *P)
{
    Float denv;
    Float envp_lln;
    Float envp_alpha;
    Float sum = 0.0;
    Int i;
}

```



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```

D->env_flg = 0;
denv = D->envlp - VT_av;
if (denv > 0) { /* drop condition */
    D->envlp = VT_av;
    D->env_drop_flg = 1;
} else {
    envlp_lin = D->envlp * P->NP->env_rate;
    envlp_alpha = D->envlp * P->NP->alpha_env * P->NP->beta_env * VT_av;
    D->envlp = envlp_lin + envlp_alpha * envlp_lin + envlp_alpha;
    if (D->env_drop_flg == 1)
        D->env_flg = 1;
    D->env_drop_flg = 0;
}

/* conditional update noise-spect (envlp) and related variables */
if (D->env_flg == 1 &&
    (D->envlp > 2. * D->M_pwr) || (D->envlp < D->M_pwr * 0.5) ) {
    if (P->DWLP_FLG == 2) { /* matching to mean spectrum only */
        if (D->ST_av < 1.414 * D->VT0_av + 1.66 /* -- 1.5dB range */
            || D->ST_av > D->VT0_av + 0.707216 ) {
            for (i = 0; i < P->win_len/2 + 1; i++) /* sample mean spec */
                D->lambdaD[i] = P->noise_bias + D->ST[i];
            D->updn_flg = 3; /* lock to mean (ST) (indicator) */
            D->VT_flg = 0;
        } else {
            if (D->envlp > 2. * D->M_pwr) { /* increase */
                if (D->ST_av > 1.414 * D->VT0_av + 1.66 /* -- 1.5dB range */
                    || D->ST_av < D->VT0_av + 0.707216 ) {
                    for (i = 0; i < P->win_len/2 + 1; i++) {
                        D->lambdaD[i] = sqrt((D->M_pwr / D->envlp) * D->VT_anth[i])
                    }
                }
            }
            D->lambdaD[i] = sqrt((D->M_pwr / D->envlp) * D->VT_anth[i])
        }
    }
    D->VT_flg = 1;
    D->updn_flg = 2; /* up to VT_anth (indicator) */
} else {
    for (i = 0; i < P->win_len/2 + 1; i++) /* sample mean spec */
        D->lambdaD[i] = D->ST[i];
    D->updn_flg = 1;
    D->VT_flg = 0;
}
for (i = 0; i < P->win_len/2 + 1; i++) /* mitt by lin factor */

```

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```

Float sum, sum2, temp, alphaDn, betaDn, *fp;
int i, count;

sum = D->lambdaD[0] + D->lambdaD[P->win_len/2];
for ( i = 1; i < P->win_len/2; i++ )
    sum += D->lambdaD[i] + D->lambdaD[i+1];

D->M_per0 = sum + P->win_len_inv;

if ( !n_flag ) { ID->n_c0 == 0 } { /* NOISE ONLY */
    temp = gamma_av * P->gN;
    betaDn = P->NP->f_n_max * ( temp > 0 ? temp : -temp );

    if ( betaDn > P->NP->f_n_max )
        betaDn = P->NP->f_n_max;
    if ( betaDn < P->NP->f_n_min )
        betaDn = P->NP->f_n_min;
    alphaDn = 1. - betaDn;

    for ( i = 0; i < P->win_len/2 + 1; i++ )
        D->lambdaD[i] = alphaDn * D->lambdaD[i] + P->noise_bias * betaDn * D->YY[i]
} else {
    if ( P->MS_FLO ) {
        temp = P->NP->f_n_min;
        sum = 0.0;
        count = 0;

        if ( D->gamak[0] < P->NP->gammas_thr ) {
            sum += D->gamak[0];
            count++;
        }

        if ( D->gamak[P->win_len/2] < P->NP->gammas_thr ) {
            sum += D->gamak[P->win_len/2];
            count++;
        }

        for ( i = 1; i < P->win_len/2; i++ ) {
            if ( D->gamak[i] < P->NP->gammas_thr ) {
                sum += D->gamak[i];
                count++;
            }
        }

        sum2 = sum / count - P->gN;

        if ( count > 0 )
            temp = P->NP->f_n_max * ( sum2 > 0 ? sum2 : -sum2 );

        if ( temp > P->NP->f_n_max )
            temp = P->NP->f_n_max;
        if ( temp < P->NP->f_n_min )
            temp = P->NP->f_n_min;

        for ( i = 0; i < P->win_len/2 + 1; i++ )
            if ( D->gamak[i] < P->NP->gammas_thr )
                D->lambdaD[i] += ( temp * D->gk[i] * ( P->noise_bias * D->YY[i] - D->
                lambdaD[i] ) );
    }
}

if ( P->ENVUP_FLO ) { /* Smoothing YY */
    if ( D->env_drop_flg == 1 ) {
        fp = D->tempV1;
        for ( i = 0; i < P->win_len/2 + 1; i++, fp += 2 ) {
            fp[0] = D->YY[i];
            fp[1] = 0.0;
        }

        if ( D->tempV2, D->tempV1, &D->fcache );
        smoothing_win ( D->tempV2, P );

        if ( D->tempV1, D->tempV2, &D->fcache );
        fp = D->tempV1;
        for ( i = 0; i < P->win_len/2 + 1; i++, fp += 2 )
            D->YY_synth[i] = fp[0];
    }
}

int i;
Float smoothed_av, alphacorr_new, alpha_M_min, alpha_M_min_1, alpha_M_min_2, alpha
_num, temp;

smoothed_av = D->smoothedspect[0]; D->smoothedspect[P->vec_lenf-1]
+ 2*vec_sumID->smoothedspect[1]; P->vec_lenf-2);
smoothed_av = smoothed_av + P->win_len_inv;
alphacorr_new = smoothed_av / YY_av - 1.0;
alphacorr_new = 1.0 / (1.0 + alphacorr_new * alphacorr_new);
D->alphacorr = 0.7 * alphacorr + 0.3 * (alphacorr_new > 0.7 ? alphacorr_new : 0.7
);

if ( USEDDOUBLES
    alpha_M_min_1 = powID->EN_LT, (P->NP->Pdecay_num));
    else
        alpha_M_min_1 = powf(D->EN_LT, (P->NP->Pdecay_num));
    sendf(
        alpha_M_min_3 = (0.3 < alpha_M_min_1 ? 0.3 : alpha_M_min_1);
        alpha_M_min = (alpha_M_min_3 < 0.05 ? 0.05 : alpha_M_min_2);
        alpha_num = P->NP->alpha_M_max * D->alphacorr;

        for ( i = 0; i < P->vec_lenf; i++ ) {
            temp = (D->smoothedspect[i] / D->noisespect[i]) * i;
            D->alpha_var[i] = alpha_num / (1+temp * temp);
            D->alpha_var[i] = (D->alpha_var[i] < alpha_M_min ? alpha_M_min : D->alpha_var[i])
        }
        D->smoothedspect[i] = D->alpha_var[i] * D->smoothedspect[i] + (1-D->alpha_var[i])
        * D->YY[i];
    }
}

```

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return ID->smoothedspect();

.....
Subroutine normalised variances, compute variance of smoothed
periodogram, normalise it, and use it to compute a
biased smoothed periodogram
.....
//get bias compensation(Enhance_Data "D, Enhance_Param "P) {

int i;
float tmp1, tmp2, beta_var, var_sp, var_rel_av_sqrt;

for (i = 0; i < P->vec_len; i++) {
    tmp1 = D->alpha_var[i] * P->alpha_var[i];
    beta_var = (tmp1 > 0.8 ? 0.8 : tmp1);
    tmp2 = (1 - beta_var) * D->smoothedspect[i];

    var_sp_av[i] = beta_var * D->var_sp_av[i] + tmp2;
    var_sp = D->var_sp_2[i] - D->var_sp_av[i] * D->var_sp_av[i];
    tmp1 = var_sp / (D->noisespect[i] * D->noisespect[i]);
    D->var_rel[i] = (tmp1 > 0.5 ? 0.5 : tmp1);
}

D->var_rel_av = (D->var_rel[0] + D->var_rel[P->vec_len-1]
+ 2 * vec_sum((D->var_rel[1], P->vec_len-2)) * P->vec_len_inv,
D->var_rel_av = (D->var_rel_av < 0 ? 0 : D->var_rel_av);

//def US2000B25
var_rel_av_sqrt = 1.0 + 1.5 * sqrt(D->var_rel_av);
//
var_rel_av_sqrt = 1.0 + 1.5 * sqrt((D->var_rel_av);
//
var_rel_av_sqrt = 1.0 + 1.5 * sqrt((D->var_rel_av);
//
tmp1 = var_rel_av_sqrt * P->NP->Pvar;
tmp2 = var_rel_av_sqrt * P->NP->Pvar;
tmp3 = var_rel_av_sqrt * P->NP->Pvar;
for (i = 0; i < P->vec_len; i++) {
    D->biased_smoothedspect[i] = (var_rel_av_sqrt * D->var_rel[i]
+ tmp1 / (P->NP->minv - D->var_rel[i])) * D->smoothedspect[i];
    biased_smoothedspect_sub[i] = (var_rel_av_sqrt * D->var_rel[i]
+ tmp2 / (P->NP->minv - D->var_rel[i])) * D->smoothedspect[i];
    biased_smoothedspect_sub[i] = (var_rel_av_sqrt * D->var_rel[i]
+ tmp3 / (P->NP->minv - D->var_rel[i])) * D->smoothedspect[i];
}

return ID->biased_smoothedspect();

.....
Subroutine noise slope: compute maximum of the permitted increase of
the noise estimate as a function of the mean signal variance
.....
//get noise slope(Enhance_Data "D, Enhance_Param "P) {

float noise_slope_max;

if (D->var_rel_av > 0.18)
    noise_slope_max = 1.2;
else if (D->var_rel_av < 0.01) || (D->vec_len < 50)
    noise_slope_max = 0;
else if (D->var_rel_av < 0.05)
    noise_slope_max = 4;
}

else if (D->var_rel_av < 0.05)
    noise_slope_max = 2;
else
    noise_slope_max = 1.2;

if (P->software_ver > 7)
    noise_slope_max = 1.1;

return (noise_slope_max);
}

//.....
Subroutine min_search: find minimum of psd's in circular buffer
//
//.....
float min_search(Enhance_Data "D, Enhance_Param "P) {

int i, k;
float noise_slope_max;

if (D->minspec_counter > P->NP->len_minwin) {
    noise_slope_max = noise_slope(D, P);
    for (i = 0; i < P->vec_len; i++) {
        if (D->biased_smoothedspect[i] < D->act_min[i]) {
            D->act_min[i] = D->biased_smoothedspect[i];
            D->act_min_sub[i] = D->biased_smoothedspect_sub[i];
            D->localflag[i] = 0;
        }
        D->circb[D->circb_inda[i]] = D->act_min[i]; // write new minimum into ring
        buffer
        D->circb_inda = ((D->circb_inda + 1) %
P->NP->num_minwin) % 0; // increment
        D->pointer++;
        D->circb_min[i] = D->circb[0]; // find minimum of ring buffer
        for (k = 1; k < P->NP->num_minwin; k++) {
            D->circb_min[i] = D->circb[k] < D->circb_min[i] ? D->circb[k] : D->circb_min[i];
            D->circb_min_sub[i] = D->circb_min_sub[i] < D->circb_min_sub[i] ? D->circb_min_sub[i] : D->circb_min_sub[i];
        }
        // rapid update in case of local minima which do not deviate more than noise_s
        //lope_max from the current minima
        if (D->localflag[i] && (D->act_min_sub[i] > D->circb_min[i]) &&
(D->act_min_sub[i] < noise_slope_max * D->circb_min[i])) {
            D->circb_min[i] = D->act_min_sub[i]; // ID->circb_min[i] < D->act_min_sub[i]
            D->circb_min_sub[i] = D->act_min_sub_sub[i]; // ID->circb_min_sub[i] < D->act_min_sub_sub[i]
        }
        D->act_min_sub[i] = D->circb_min_sub[i]; // ID->act_min_sub[i] < D->circb_min_sub[i]
    }
    // propagate new rapid update minimum into ring buffer
    for (k = 0; k < P->NP->num_minwin; k++)
        D->circb[k] = D->circb_min[k];
}
D->act_min[i] = P->vec_max; // initialise minimum with largest possible
value
D->localflag[i] = 0; // reset local minimum indicator
D->minspec_counter = 1;
}
else {
    if (D->minspec_counter > 1) {
        for (i = 0; i < P->vec_len; i++) {

```

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```

/* set up FFT cache */
fftwinit( &D->cache, 8 );

/* initialize noise spectrum */
for ( j = 0; j < P->win_len2; j++)
    D->tempV1[j] = 0.0;

fp2 = 0->init_noise;
for ( j = 0; j < P->noise_frames; j++) {
    for ( i = 0; i < P->win_len; i++) {
        D->tempV2[i] = P->analyze_window[i] * fp2[i];
    }
    fftf ( D->tempV1, D->tempV2, &D->cache );
    /* get rough estimate of lambdaD */
    D->tempV1[0] = D->tempV1[0]*D->tempV1[0] + P->win_len_inv;
    D->tempV1[1] = 0.0;
    D->tempV1[P->win_len] = D->tempV1[P->win_len]*D->tempV1[P->win_len] + P->win_len_inv;
    D->tempV1[P->win_len+1] = 0.0;
    fp = D->tempV1 + 2;
    for ( i = 2; i < P->win_len/2 + 1; i++, fp += 2 ) {
        fp[0] = ( fp[0] + fp[i] + fp[i] + fp[i] ) * P->win_len_inv;
        fp[i] = 0.0;
    }
    for ( i = 0; i < P->win_len/2; i++)
        D->tempV1[i] = D->tempV1[i];
    fp2 += P->win_len;
}

for ( j = 0; j < P->win_len2; j++)
    D->tempV1[j] = D->tempV1[j] / (float) P->noise_frames;
/* then smooth nn to get lambdaD if we didn't average */
if ( P->noise_frames == 1 ) {
    fftf ( D->tempV2, D->tempV1, &D->cache );
    smoothing_win ( D->tempV2, P );
    fftf ( D->tempV1, D->tempV2, &D->cache );
    D->lambdaD[0] = P->NP->noise_b_by_nsmth_b + D->tempV1[0] + 0.001;
    D->lambdaD[P->win_len/2] = P->NP->noise_b_by_nsmth_b + D->tempV1[P->win_len] + 0.001;
    sum = D->lambdaD[0] + D->lambdaD[P->win_len/2];
    fp = D->tempV1 + 2;
    for ( i = 1; i < P->win_len/2; i++, fp += 2 ) {
        D->lambdaD[i] = P->NP->noise_b_by_nsmth_b + fp[0] + 0.001;
        sum += D->lambdaD[i] + D->lambdaD[i];
    }
} else {
    D->lambdaD[0] = P->noise_bias + D->tempV1[0] + 0.001;
}

```

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```

D->lambda0[P->win_len/2] = P->noise_bias + D->temp[P->win_len] + 0.001;

sum = D->lambda0[0] + D->lambda0[P->win_len/2];

for (p = D->temp[0]; p < P->win_len/2; p++) {
    for (i = 1; i < P->win_len/2; i++) {
        D->lambda0[i] = P->noise_bias + p[0] + 0.001;
        sum += D->lambda0[i] + D->lambda0[i+1];
    }
}

D->M_pwr = sum + P->win_len_inv;

for (i = 0; i < P->win_len/2 + 1; i++) {
    D->v[0][i] = D->lambda0[i];
    D->v[1][i] = 0.0;
}

v_cfl[D->v[0], 0.5, P->vec_lenf];

/*def MALAM */ /* Initializations for Malah's noise estimator */
D->v[1][0] = CALLOC_FLOATIP->vec_lenf;
D->v[1][1] = CALLOC_FLOATIP->vec_lenf;

for (i = 0; i < P->win_len/2 + 1; i++)
    D->v[1][i] = D->lambda0[i];

/* Set CH_min and kst_min */
D->M_pwr0 = D->M_pwr;

/*def USEDoubles */
D->M_pwr = 10 * log10(1 + D->M_pwr);
D->ndiff = D->M_pwr - P->M_pwr;
D->CH_min = pow(10, -D->ndiff / 20.0);
/*
D->M_pwr = 10 * log10(1 + D->M_pwr);
D->ndiff = D->M_pwr - P->M_pwr;
D->CH_min = pow(10, -D->ndiff / 20.0);
endif

D->CH_min > P->M_pwr - CH_MIN
D->CH_min = P->M_pwr - CH_MIN;

D->kst_min = D->CH_min;
if (D->kst_min < 0.1)
    D->kst_min = 0.1;
if (D->kst_min > P->M_pwr - CH_MIN)
    D->kst_min = P->M_pwr - CH_MIN;
if (D->CH_min < D->kst_min)
    D->CH_min = D->kst_min;

D->CH_min = D->CH_min;
D->envp = D->M_pwr;
D->env_flg = 0;
D->env_drop_flg = 0;

/*updn_flg = 0;
D->v[1][0] = 0;

log /* Initializations for the minimum statistics noise estimator */
minstat_init(D,P);

```

```

vec_mult(D->temp, P->analysis_window, P->analysis_window, P->win_len);
/* compute initial long term SNR; speech signal power depends on the window;
the Hanning window is used as a reference here with a squared norm of 96 */
D->SN_L7 = 1.046 / D->M_pwr + vec_sum(D->temp, P->win_len) / (96.0 * P->win_ratio);

D->SN_L70 = D->SN_L7;
D->v[1][0] = 0.;
D->kst_min_var = D->kst_min;

}

/*def MINSTAT */
/*.....
/* Subroutine minstat_init: Initialization of variables for minimum
/* statistics noise estimation
/*.....
void minstat_init(Enhance_Data *D, Enhance_Params *P)
{
    /* Initialize Minimum Statistics Noise Estimator */
    int i, k;

    D->smoothespect = CALLOC_FLOATIP->vec_lenf;
    D->biased_smoothespect = CALLOC_FLOATIP->vec_lenf;
    D->circb_min = CALLOC_FLOATIP->vec_lenf;
    D->act_min = CALLOC_FLOATIP->vec_lenf;
    D->act_min_sub = CALLOC_FLOATIP->vec_lenf;
    D->noisespect = CALLOC_FLOATIP->vec_lenf;
    D->alpha_var = CALLOC_FLOATIP->vec_lenf;
    D->circb = CALLOC_FLOATIP->M_pwr - num_minwin; /* ring buffer */

    for (i=0; i < P->num_minwin; i++) {
        D->circb[i] = CALLOC_FLOATIP->vec_lenf;
        for (k = 0; k < P->vec_lenf; k++) {
            D->circb[i][k] = D->lambda0[k] * P->M_pwr;
        }
    }
    D->circb_index = 0; /* ring buffer pointer */

    D->localflag = CALLOC_SHORTIP->vec_lenf;
    D->minspec_counter = P->M_pwr - len_minwin;

    D->var_sp_av = CALLOC_FLOATIP->vec_lenf;
    D->var_sp_2 = CALLOC_FLOATIP->vec_lenf;
    D->var_rai = CALLOC_FLOATIP->vec_lenf;

    for (k = 0; k < P->vec_lenf; k++) {
        D->smoothespect[k] = D->lambda0[k] * P->M_pwr;
        D->act_min[k] = D->lambda0[k] * P->M_pwr;
        D->act_min_sub[k] = D->lambda0[k] * P->M_pwr;
        D->noisespect[k] = D->lambda0[k] * P->M_pwr;
        D->circb_min[k] = D->lambda0[k] * P->M_pwr;

        D->var_sp_av[k] = 1.2247448713159 * D->noisespect[k]; /* sqrt(3/2) */
        D->var_sp_2[k] = 2 * D->noisespect[k] + D->noisespect[k];
    }
    D->alphacorr = 0.9;
}

```


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enh_fun.c.

```

/* Set CH_min and Kst_min */
#define USDOURSIZE
D->n_per = 10 * log10(1 + D->n_per);
D->ndiff = D->n_per - P->np->resn_thr;
D->CH_min = pow(10, -D->ndiff / 20.0);
/*
D->n_per = 10 * log10(1 + D->n_per);
D->ndiff = D->n_per - P->np->resn_thr;
D->CH_min = pow(10, -D->ndiff / 20.0);
*/
if (D->CH_min > P->np->CH_MIN)
    D->CH_min = P->np->CH_MIN;

D->Kst_min = D->CH_min * P->rate_factor;
/* ID->Kst_min = 0.1;
   ID->Kst_min = 0.1;
   D->Kst_min = P->np->CH_MIN;
   D->Kst_min = P->np->CH_MIN;
   D->Kst_min = D->Kst_min;
*/

.....
/* Subroutine minstet_terminate: Terminate execution of noise minimum
   statistics noise estimator
   .....
   minstet_terminate(Enhance_Data *D, Enhance_Params *P)
   .....
   int i, n_flag;
   float sum = 0.0;
   float *fp;
   float Yt_av, gamma_av, gamma_max;
   D->flag;

   /* analysis window */
   vec_mult(D->analy, P->analysis_window, inspeech, P->win_len);

   /* into frequency domain - D->Y, D->YD->Y, Yt_av, real and imaginary parts are int
   released */
   if (D->Y, D->analy, AD->cache);

   D->Y[0] = D->Y[0] * D->Y[0];
   D->Y[P->vec_len-1] = D->Y[P->win_len] * D->Y[P->win_len];
   vec_mag(D->Y[1], AD->Y[2], P->vec_len-1);
   vec_sqrt(D->Y[2], D->Y, P->vec_len);
   vec_mult_const(D->Y, D->Y, P->win_len_inv, P->vec_len);
   vec_add_const(D->Y, D->Y, 0.001, P->vec_len);

   if (P->slphay != 0.)
       for (i = 0; i < P->win_len/2 + 1; i++)
           D->Y[i] = P->slphay * D->Y[i] + P->slphay * D->Y[i];
   sum = 3*vec_sum(D->Y[1], P->vec_len-1) * D->Y[0] * D->Y[P->vec_len-1];
   Yt_av = sum * P->win_len_inv;

.....
/* Subroutine minstet_terminate: Terminate execution of noise minimum
   statistics noise estimator
   .....
   minstet_terminate(Enhance_Data *D, Enhance_Params *P)
   .....
   int i;

   free(D->smoothspect);
   free(D->biased_smoothspect);
   free(D->biased_smoothspect_sub);
   free(D->clircb_min);
   free(D->sect_min);
   free(D->sect_min_sub);
   free(D->noisespect);
   free(D->clircb);
   free(D->clircb);
   free(D->localflag);
   free(D->var_ep_av);
   free(D->var_ep_2);
   free(D->var_ell);

.....
/* Subroutine enh_terminate: Terminate execution of program
   .....
   enh_terminate(Enhance_Data *D, Enhance_Params *P)
   .....
   #define MAXLEN
   P->P;
   free(D->Yt_maxh);
   free(D->Yt);

```



enh_fun.c.

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```

/* computation of lower envelope and setting env flags */
/* if (P->ENVLP_FLAG)
   track_envelope(YT_av, D, P); */

#define MINSTAY

/* compute smoothed short time periodogram */
smoothed_periodogram(D, YT_av, P);

/* compute inverse bias and multiply short time periodogram with inverse bias */
bias_compensation(D, P);

/* determine unbiased noise and estimate by minimum search */
min_search(D, P);

/* compute 'gamma' */
vec_div(D->gamma, D->YT, D->lambdaD, P->vec_lenf);
gamma_max = vec_max(D->gamma, P->vec_lenf);
sum = D->gamma[0] + D->gamma[P->vec_lenf-1] + 2 * vec_sum(D->gamma[1], P->vec_lenf);

gamma_av = sum * P->win_len_inv;

/* determine signal presence */
n_flag = 0; /* default flag - signal present */
if (gamma_max < P->NP-gamma_thr) {
    n_flag = 1; /* noise-only */
    if (YT_av > D->M_per * P->NP-gamma_thr + 3.)
        n_flag = 0; /* overriding if frame SNR > 3dB (9/98) */
}

if (D->sl == 1) {
    /* initial estimation of a priori SNR and Gain */
    n_flag = 1;
    for (i = 0; i < P->vec_lenf; i++) {
        D->sl[i] = D->sl_min;
        D->sl[i] = P->sl_max;
        D->Gain[i] = D->Gain_min;
        D->CH[i] = D->CH_min;
        D->GainD[i] = D->GainD_min;
        D->sl[i] = D->Ymag[i] * D->CH_min;
    }
} else { /* D > 1 */
    /* estimation of a priori SNR */
    for (i = 0; i < P->vec_lenf; i++) {
        D->sl[i] = P->slphak * D->sl[i] * D->sl_inv / D->lambda
            + P->slphak * (D->gamma[i] > P->slphak ? D->gamma[i] * P->slphak : 0);
    }
    D->sl_min_var = 0.9 * D->sl_min_var + 0.1 * sl_min_adapt(n_flag, D->sl_min, D->vec_limit_bottom(D->sl, D->sl_min_var, P->vec_lenf));
    /* estimation of k-th component 'signal absence' prob and gain */
    vec_fill(D->sl, P->sl_max, P->vec_lenf); /* default value for qk's (9/98)

```


vect_fun.h.

```

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#define __vect_fun__
#define __vect_fun__
.....
vect_fun.h - Functions for MATLAB - like vector operations
Author: Rainer Martin, AT&T Labs-Research
Last Update: $Id$
.....
include "globals.h"

void float_to_short(float input[], short output[], int num_samples);

void vec_copy(float *vec1, float *vec2, int m);
void vec_accum(float *vec1, float *vec2, int m);
void vec_add(float *vec1, float *vec2, float *vec3, int m);
void vec_mult(float *vec1, float *vec2, float *vec3, int m);
void vec_mult_const(float *vec, float *vec2, float c, int m);
void vec_div(float *vec1, float *vec2, float *vec3, int m);
void vec_inv(float *vec1, float *vec2, float *vec3, int m);
void vec_neg(float *vec1, float *vec2, float *vec3, int m);
void vec_sum(float *vec, int m);
void vec_add_const(float *vec1, float *vec2, float c, int m);
void vec_max(float *vec, int m);
void vec_min(float *vec, int m);
void vec_sqrt(float *vec1, float *vec2, int m);
void vec_limit_bottom(float *vec1, float *vec2, float c, int m);
void vec_limit_top(float *vec1, float *vec2, float c, int m);
void vec_fill(float *vec, float c, int m);
endif

```



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vect_fun.c

```

#include "vect_fun.h"

.....
vect_fun.c - Functions for MATLAB - like vector operations
Author: Rainer Martin, AT&T Labs-Research
Last Update: $Id$
.....

.....
Subroutine (float_to_short): round float samples to short samples
.....
st_to_short(float input[], short output[], int num_samples)
int i;
for (i = 0; i < num_samples; i++) {
    if (input[i] > 32767.)
        output[i] = 32767;
    else if (input[i] < -32768.)
        output[i] = -32768;
    else if (input[i] >= 0.)
        output[i] = (short) (input[i] * .5);
    else
        output[i] = (short) (input[i] * -.5);
}

.....
Subroutine vec_copy: copy vector vec2 of float samples into vector vec1
.....
float "vec_copy(float "vec1, float "vec2, int m)
int i;
for (i = 0; i < m; i++)
    vec1[i] = vec2[i];
return(vec1);

.....
Subroutine vec_accu: add m samples of vec2 to vec1
.....
float "vec_accu(float "vec1, float "vec2, int m)
int i;
for (i = 0; i < m; i++)
    vec1[i] += vec2[i];
return(vec1);

.....
Subroutine vec_add: add m samples of vec2 to vec1, store result in vec1
.....
float "vec_add(float "vec1, float "vec2, float "vec3, int m)
int i;
for (i = 0; i < m; i++)
    vec3[i] = vec1[i] + vec2[i];
return(vec3);

.....
Subroutine vec_mag2: Y is magnitude squared of vector Y.
.....
float "vec_mag2(float "Y, float "Y, int m)
int i;
float "fp;
fp = Y;
for (i = 0; i < m; i++) { fp += 2;
    Y[i] = (fp) * (fp) + fp[i] * (fp[i]);
}
return(Y);

.....
Subroutine vec_sum: computes the sum of vector components.
.....
float vec_sum(float "vec, int m)

```

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vect_fun.c.

```

Float tmp;
int i;

tmp = 0;
for (i = 0; i < m; i++)
    tmp += vec[i];
return(tmp);

.....
Subroutine vec_mult_const: Multiply m samples with constant.
.....
vec[i] = vec2[i] * c
.....
ec_mult_const(Float 'vec1, Float 'vec2, Float c, int m)
int i;
for (i = 0; i < m; i++)
    vec[i] = vec2[i] * c;
return(vec);

.....
Subroutine vec_add_const: add constant c to m samples.
.....
vec[i] = vec2[i] + c
.....
vec_add_const(Float 'vec1, Float 'vec2, Float c, int m)
int i;
for (i = 0; i < m; i++)
    vec[i] = vec2[i] + c;
return(vec);

.....
Subroutine vec_limit_bottom: compute vec2[i] with a constant c and take
.....
minimum.
.....
Float 'vec_limit_bottom(Float 'vec1, Float 'vec2, Float c, int m)
int i;
for (i = 0; i < m; i++)
    vec[i] = (vec2[i] < c) ? c : vec2[i];
return(vec);

.....
Subroutine vec_limit_top: compute vec2[i] with a constant c and take
.....
minimum.
.....
Float 'vec_limit_top(Float 'vec1, Float 'vec2, Float c, int m)
int i;
for (i = 0; i < m; i++)
    vec[i] = (vec2[i] > c) ? c : vec2[i];
return(vec);

.....
Subroutine vec_min: computes the minimum of vector components.
.....
Float vec_min(Float 'vec, int m)
int i;
tmp = vec[0];
for (i = 1; i < m; i++)
    if (vec[i] < tmp)
        tmp = vec[i];
return(tmp);

.....
Subroutine vec_max: computes the maximum of vector components.
.....
Float vec_max(Float 'vec, int m)
int i;
tmp = vec[0];
for (i = 1; i < m; i++)
    if (vec[i] > tmp)
        tmp = vec[i];
return(tmp);

.....
Subroutine vec_sqrt: compute sqrt of vec2. vec[i] = sqrt(vec2[i])
.....
vec_sqrt(Float 'vec1, Float 'vec2, int m)
int i;
for (i = 0; i < m; i++)
    vec[i] = sqrt(vec2[i]);
return(vec);

.....
Subroutine vec_min: computes the minimum of m vector components.
.....
Float 'vec_min(Float 'vec, int m)
int i;
tmp = vec[0];
for (i = 1; i < m; i++)
    if (vec[i] < tmp)
        tmp = vec[i];
return(tmp);

.....
Subroutine vec_max: computes the maximum of m vector components.
.....
Float 'vec_max(Float 'vec, int m)
int i;
tmp = vec[0];
for (i = 1; i < m; i++)
    if (vec[i] > tmp)
        tmp = vec[i];
return(tmp);

```

3

vect_fun.c.

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for(i=0; i < m; i++)
 vec[i] = c;
return(vec);

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ffreal.c.

..... ffrreal.c
.....

Fast FFT of a real time sequence based on viewing the full-size real-data transform as an half-size complex-data transform.

ffr.c is the inverse of ftr.c. The two routines are complementary in the sense that the constants in w[i][2] and b[i] are the same and may be initialized only once by either routine.

T. Shoham 3/95 94/95 p. 378

Modified by D. A. Kapilow 7/95, to speed it up on both the PC and S32.

n/ *enhance.h*

..... Bit reversal function
.....

n 32-bit input integer

ndim Power-of-2 number. Log2(ndim) is the

number of LS bits from "n" to reverse. The operation is:

$$b[i] = b[\log_2(nbit) - i], \quad i=0, \log_2(ndim) - 1$$

where b[i] is the bit at location i.

The function returns an integer whose nbit LS-bits are the reversal of same bits in "n". Other bits are 0.

```
/*c int brr(int n, int ndim)
```

```
int j,m,k;
```

```
m = 0;
```

```
j = 1;
```

```
for (k = ndim >> 1; k > 0; j <<= 1,k >>= 1)
```

```
if (j & k)
```

```
m = m | k;
```

```
return(m);
```

Allocate and initialize the constant data for the FFT */

```
.. ffrinit
```

```
fftr *fb, /* out: initialize it */
```

```
int *nsize /* in: initialize */
```

```
int i, nsize, *br,
```

```
float t, pm, *w;
```

```
(b->size = 1) <<= initialize;
```

```
b->size = (b->size >> 1);
```

/* allocate coefficient and work arrays */

```
if ((b->coeffs = (float *) malloc(sizeof(float) * (b->size))) == NULL ||
```

```
(b->br = (int *) malloc(sizeof(int) * nsize)) == NULL)
```

```
fprintf(stderr, "malloc error\n");
```

```
w = (float *) malloc(2);
```

```
br = (b->br);
```

```
pm = 2. * PI / (b->size);
```

```
for (i = 1; i <= nsize; i++) {
```

```
t = pm * i;
```

```
w[i] = cos(t);
```

```
w[i] = -sin(t);
w++ 2;
}
for(i = 0; i <= nsize; i++)
br[i] = brr(i, nsize) << 1;
fb->nsize = 1. / (b->size);
```

```
/* Free the cached data */
```

```
void ffrdone(fftr *fb)
```

```
{
```

```
free(fb->coeffs);
```

```
free(fb->br);
```

```
fb->coeffs = 0;
```

```
fb->br = 0;
```

```
}
```

```
void ffrtr
```

```
float *y, /* out: complex FFT */
```

```
float *x, /* in: real signal */
```

```
fftr *fb /* in: cached FFT parameters */
```

```
{
```

```
int i, j, k, l, winc, ka, nsize, nsizeq;
```

```
int *br;
```

```
float t0, t1, u0, u1, w0, w1, pm;
```

```
float *y0p, *y1p, *w0p, *w1;
```

```
/* Initialization */
```

```
nsize = fb->nsize;
```

```
br = fb->br;
```

```
w = fb->coeffs;
```

```
nsizeq = nsize >> 1;
```

```
nsizeq = nsize >> 2;
```

/* Make the full-size real input an half-size complex */

```
y0p = y;
```

```
for(k = 0; k <= nsizeq; k++) {
```

```
y1p = x[br[k]]; /* y1p = x[br[k]] */
```

```
y0p[0] = y1p[0];
```

```
y0p[1] = y1p[1];
```

```
y0p++ 2;
```

```
}
```

/* Half-size complex FFT */

```
/* 1st stage nsize/2 simple butterflies */
```

```
y0p = y;
```

```
for(k = 0; k <= nsizeq; k++) {
```

```
t0 = y0p[0];
```

```
t1 = y0p[1];
```

```
u0 = y0p[2];
```

```
u1 = y0p[3];
```

```
y0p[0] = t0 + u0;
```

```
y0p[1] = t1 + u1;
```

```
y0p[2] = t0 - u0;
```

```
y0p[3] = t1 - u1;
```

```
y0p++ 4;
```

```
}
```

/* Next stages */

```
for (l = 2; winc = nsize, l <= nsizeq; winc >>= 1) {
```

```
ka = l - 1;
```

```
l <<= 1;
```

```
l <= l << 1;
```

```
for (j = 0; j <= nsizeq; j++ 1) {
```

```
y0p = y[br[j]]; /* y0p = y[br[j]] */
```

```
y1p = y0p + 1;
```


ffreal.c.

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```

/* Half-size inverse FFT */
/* Do nalist/2 simple butterflies */
y0p = y;
for (k = 0; k < nalist; k++) {
    t0 = y0p[0];
    t1 = y0p[1];
    u0 = y0p[2];
    u1 = y0p[3];
    y0p[0] = t0 + u0;
    y0p[1] = t1 + u1;
    y0p[2] = t0 - u0;
    y0p[3] = t1 - u1;
    y0p += 4;
}

/* Next stages */
for (i = 2; winc = nalist; i < nalist; winc <<= 1) {
    ks = 1 - i;
    j = ks + 1;
    for (j = 0; j < nalist; j += 1) {
        y0p = y[j];
        y1p = y0p + i;
        wp = &winc;
        t0 = y0p[0];
        t1 = y0p[1];
        u0 = y1p[0];
        u1 = y1p[1];
        y0p[0] = t0 + u0;
        y0p[1] = t1 + u1;
        y1p[0] = t0 - u0;
        y1p[1] = t1 - u1;
        for (k = 0; k < ks; k++) {
            y0p += 2;
            y1p += 2;
            u0 = y0p[0] + y1p[0] + wp[1] + y1p[1];
            u1 = y0p[0] + y1p[1] - wp[1] + y1p[0];
            t0 = y0p[0];
            t1 = y0p[1];
            y0p[0] = t0 + u0;
            y0p[1] = t1 + u1;
            y1p[0] = t0 - u0;
            y1p[1] = t1 - u1;
            wp += winc;
        }
    }
}

/* Scale it */
t0 = fb->inverse;
for (k = 0; k < nalist; k++)
    y[k] *= t0;

```

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ffreal.h.

..... ff1.c

Fast FFT of a real time sequence based on viewing
the full-size real-data transform as an half-size
complex-data transform.

T. Shoham 5/95 94/95 p. 178

Modified by D. A. Kaplow 7/95, to speed it up on both the PC
and SGI.

typedef struct fffcache

int nsize; /* size of the FFT - power of 2 */
int *br; /* N/2 reversal index array of size nsize/2 */
float *cosin; /* Complex FFT constants of size nsize/2 */
float *invtab; /* 1. / size */

};

id fffr(float*, float*, fffcache);
id ifffr(float*, float*, fffcache);
id fffinit(fffcache*, int);
id fffdone(fffcache*);

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16:00:54

globals.h.

```

#include _globals_
file _globals_
.....
globals.h - Compilation Switches and Constants
Author: Rainer Martin, AT&T Labs-Research
Last Update: 8/01/8
.....

#include <stdio.h>
#include <stdlib.h>
#include <math.h>
/*
 * precision: choice of USEDoubles or USEFloats */
#define USEDoubles

define the type of noise estimator to be used.
MINSTAT is the optimal smoothing minimum statistics estimator
MALAN is David Malah's noise estimation method. This is not fully implemented. */
#define MINSTAT /* choice of MALAN or MINSTAT */

define the file format for the enhanced speech: WRITEFloat writes the data
in the float format which might be actually double or float.
WRITEShort writes 16 bit short data. */
#define WRITEShort /* choice of WRITEShort or WRITEFloat */

.....
CONSTANTS .....
.....
#define PI 3.14159265358979323846

#define USEDoubles
#define double Float;
.....
#define USEFloats
#define float Float;
.....
#define short Word16;
#define long Word32;
.....
#endif

```

windows.h.

2/05/99
16:01:08

```

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windows.h.

c2/05/93
16:01:08

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46) 0

```

nd1f

windows.h.



WHAT IS CLAIMED IS:

1. A method for enhancing a speech signal for use in speech coding, the speech signal representing background noise and periods of articulated speech, the speech signal being divided into a plurality of data frames, the method comprising the steps of:

5 applying a transform to the speech signal of a data frame to generate a plurality of sub-band speech signals;

 making a determination whether the speech signal corresponding to the data frame represents articulated speech;

10 applying individual gain values to individual sub-band speech signals, wherein the lowest permissible gain value for a frame determined to represent articulated speech is lower than the lowest permissible gain value for a frame determined to represent background noise only; and

 applying an inverse transform to the plurality of sub-band speech signals.

2. The method of claim 1 further comprising the step of determining the individual gain values and wherein the lowest permissible gain value is a function of a lowest permissible *a priori* signal to noise ratio.

3. A method for enhancing a signal for use in speech coding, the signal being divided into data frames and representing background noise information and periods of articulated speech information, the method comprising the steps of:

5 making a determination whether the signal of a data frame represents articulated speech information; and

applying a gain value to the signal, wherein the lowest permissible gain value for a frame determined to represent articulated speech is lower than the lowest permissible gain value for a frame determined to represent background noise only.

10

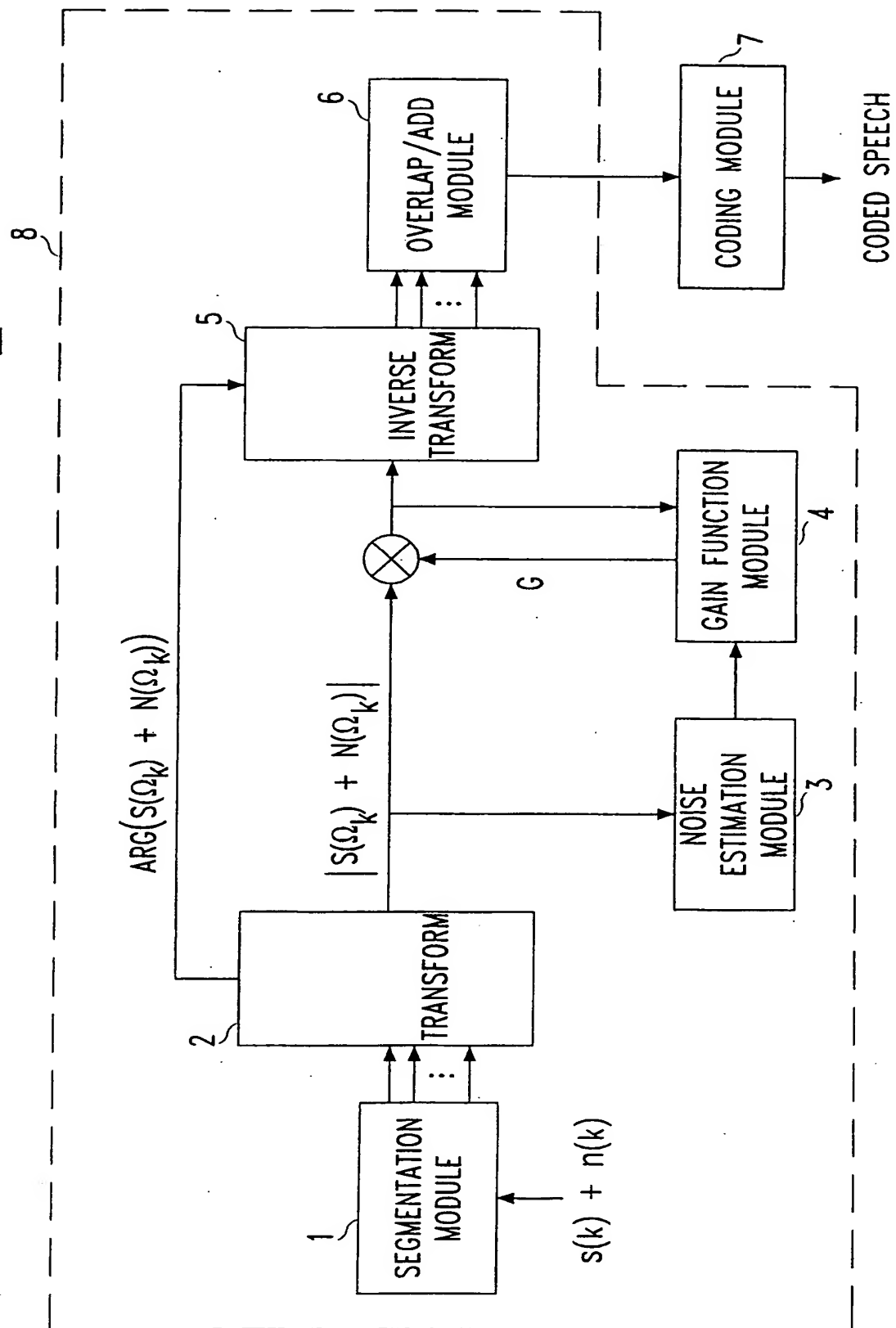
4. The method of claim 3 further comprising the step of determining the gain value and wherein the lowest permissible gain value is a function of a lowest permissible *a priori* signal to noise ratio.

5

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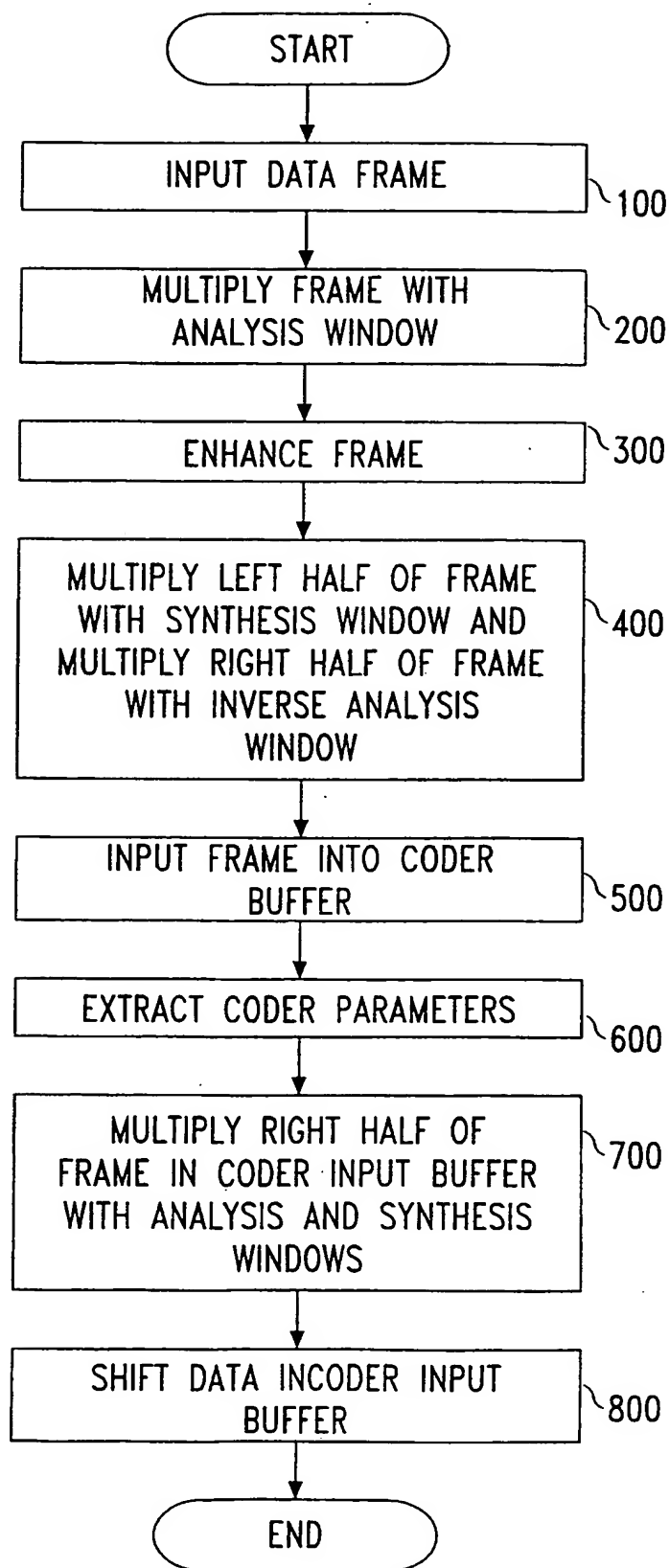
FIG. 1

10



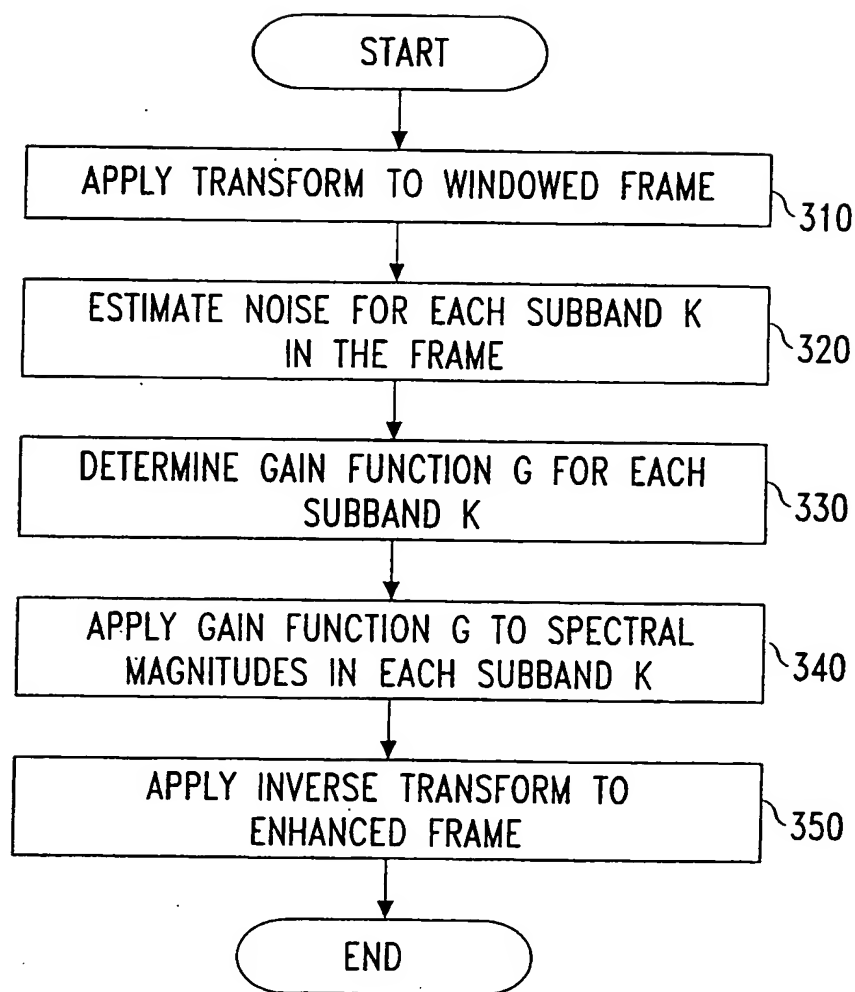
2/5

FIG. 2



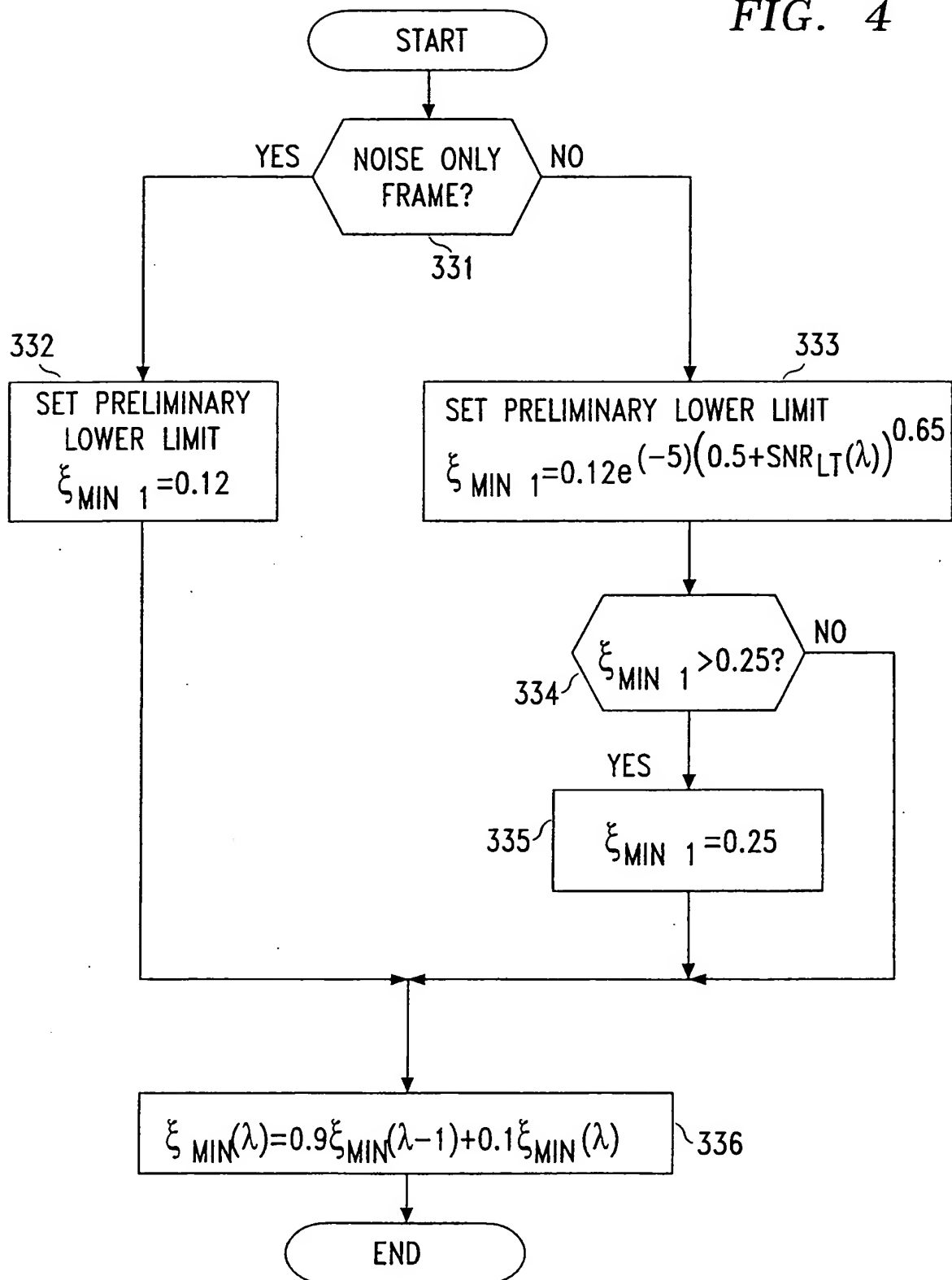
3/5

FIG. 3



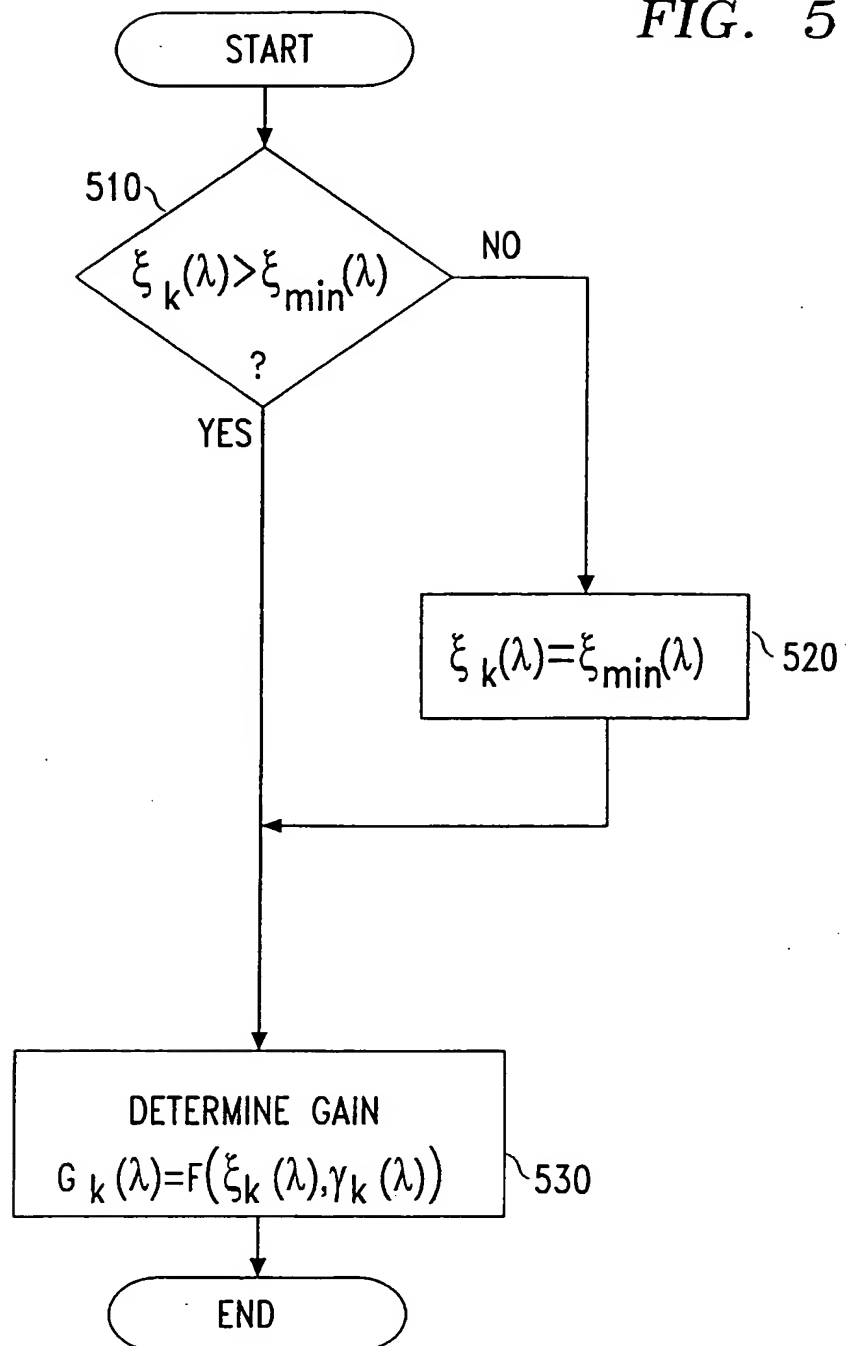
4/5

FIG. 4



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FIG. 5



INTERNATIONAL SEARCH REPORT

International Application No

PCT/US 00/03372

A. CLASSIFICATION OF SUBJECT MATTER

IPC 7 G10L21/02

According to International Patent Classification (IPC) or to both national classification and IPC

B. FIELDS SEARCHED

Minimum documentation searched (classification system followed by classification symbols)

IPC 7 G10L

Documentation searched other than minimum documentation to the extent that such documents are included in the fields searched

Electronic data base consulted during the international search (name of data base and, where practical, search terms used)

C. DOCUMENTS CONSIDERED TO BE RELEVANT

Category *	Citation of document, with indication, where appropriate, of the relevant passages	Relevant to claim No.
P, X	MARTIN R ET AL: "New speech enhancement techniques for low bit rate speech coding" 1999 IEEE WORKSHOP ON SPEECH CODING PROCEEDINGS. MODEL, CODERS, AND ERROR CRITERIA (CAT. NO.99EX351), PROCEEDINGS OF 1999 IEEE WORKSHOP ON SPEECH CODING PROCEEDINGS. MODEL, CODERS, AND ERROR CRITERIA, PORVOO, FINLAND, 20-23 JUNE 1999, pages 165-167, XP002139862 1999, Piscataway, NJ, USA, IEEE, USA ISBN: 0-7803-5651-9 paragraph '0002!	1-4
A	US 5 839 101 A (HAEKKINEN JUHA ET AL) 17 November 1998 (1998-11-17) column 9, line 5 - line 45 column 10, line 24 - line 28 --- -/--	1, 3

☒ Further documents are listed in the continuation of box C.

☒ Patent family members are listed in annex.

* Special categories of cited documents :

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- *O* document referring to an oral disclosure, use, exhibition or other means
- *P* document published prior to the international filing date but later than the priority date claimed

T later document published after the international filing date or priority date and not in conflict with the application but cited to understand the principle or theory underlying the invention

X document of particular relevance; the claimed invention cannot be considered novel or cannot be considered to involve an inventive step when the document is taken alone

Y document of particular relevance; the claimed invention cannot be considered to involve an inventive step when the document is combined with one or more other such documents, such combination being obvious to a person skilled in the art.

& document member of the same patent family

Date of the actual completion of the international search

9 June 2000

Date of mailing of the international search report

29/06/2000

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Fax: (+31-70) 340-3016

Authorized officer

Krembel, L

INTERNATIONAL SEARCH REPORT

Int. donal Application No

PCT/US 00/03372

C.(Continuation) DOCUMENTS CONSIDERED TO BE RELEVANT

Category *	Citation of document, with indication, where appropriate, of the relevant passages	Relevant to claim No.
A	US 5 012 519 A (AIZNER MENDEL ET AL) 30 April 1991 (1991-04-30) figure 13 column 10, line 18 - line 36	1,3
A	CAPPÉ O: "ELIMINATION OF THE MUSICAL NOISE PHENOMENON WITH THE EPHRAIM AND MALAH NOISE SUPPRESSOR" IEEE TRANSACTIONS ON SPEECH AND AUDIO PROCESSING, US, IEEE INC. NEW YORK, vol. 2, no. 2, 1 April 1994 (1994-04-01), pages 345-349, XP000575351 ISSN: 1063-6676 paragraph '0003!	1,3
A	SCALART P ET AL: "Speech enhancement —based on a priori signal to noise estimation" 1996 IEEE INTERNATIONAL CONFERENCE ON ACOUSTICS, SPEECH, AND SIGNAL PROCESSING CONFERENCE PROCEEDINGS (CAT. NO.96CH35903), 1996 IEEE INTERNATIONAL CONFERENCE ON ACOUSTICS, SPEECH, AND SIGNAL PROCESSING CONFERENCE PROCEEDINGS, ATLANTA, GA, USA, 7-10 M, pages 629-632 vol. 2, XP002139863 1996, New York, NY, USA, IEEE, USA ISBN: 0-7803-3192-3 page 629, column 2, line 24 -page 630, column 1, line 11	1,3

INTERNATIONAL SEARCH REPORT

Information on patent family members

International Application No

PCT/US 00/03372

Patent document cited in search report	Publication date	Patent family member(s)	Publication date
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		AU 1067797 A	03-07-1997
		AU 1067897 A	03-07-1997
		EP 0790599 A	20-08-1997
		EP 0784311 A	16-07-1997
		WO 9722116 A	19-06-1997
		WO 9722117 A	19-06-1997
		JP 9212195 A	15-08-1997
		JP 9204196 A	05-08-1997
		US 5963901 A	05-10-1999
US 5012519 A	30-04-1991	NONE	